

Intelligent Power Module (IPM)

Inverter, 1200 V, 15 A

NFAM1512L7B

General description

NFAM1512L7B is an advanced IPM module providing a fully-featured, high-performance inverter output stage for AC Induction, BLDC and PMSM motors. These modules integrate optimized gate drive of the built-in IGBTs to minimize EMI and losses, while also providing multiple on-module protection features including under-voltage lockouts, over-current shutdown, thermal monitoring of drive IC, and fault reporting. The built-in, high-speed HVIC requires only a single supply voltage and translates the incoming logic-level gate inputs to the high-voltage, high-current drive signals required to properly drive the module's internal IGBTs. Separate negative IGBT terminals are available for each phase to support the widest variety of control algorithms.

Features

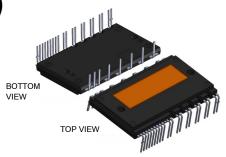
- 1200 V 15 A 3-Phase FS7 IGBT Inverter, Including Control ICs for Gate Drive and Protections
- Very Low Thermal Resistance Using Al2O3 DBC Substrate
- Active Logic Interface
- Built-in Under-voltage Protection (UVP)
- Built-In Bootstrap Diodes/Resistors
- Separate Low-side IGBT Emitter Connections for Individual Current Sensing of Each Phase

1

- Temperature Sensor (TSU Output by LVIC)
- UL Certification: E209204
- This is a Pb-Free Device

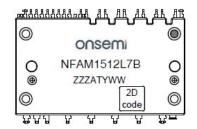
Typical Application

- · Industrial Drives
- Industrial Pumps
- Industrial Fans
- Industrial Automation



DIP39, 54.5x31.0 EP-2 CASE MODGX

MARKING DIAGRAM



NFAM1512L7B = Specific Device Code
ZZZ = Assembly Lot Code
A = Assembly Location
T = Test Location
Y = Year
WW = Work Week

ORDERING INFORMATION

Device	Package	Shipping
NFAM1512L7B	DIP39, 31.0x54.5 (Pb-Free)	90 / BOX

PIN CONFIGURATION

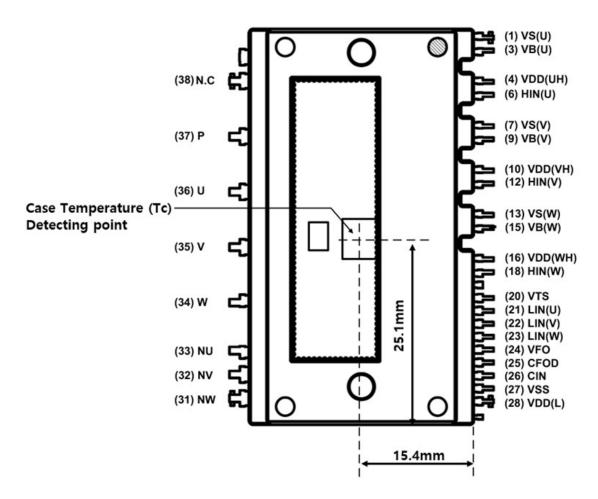


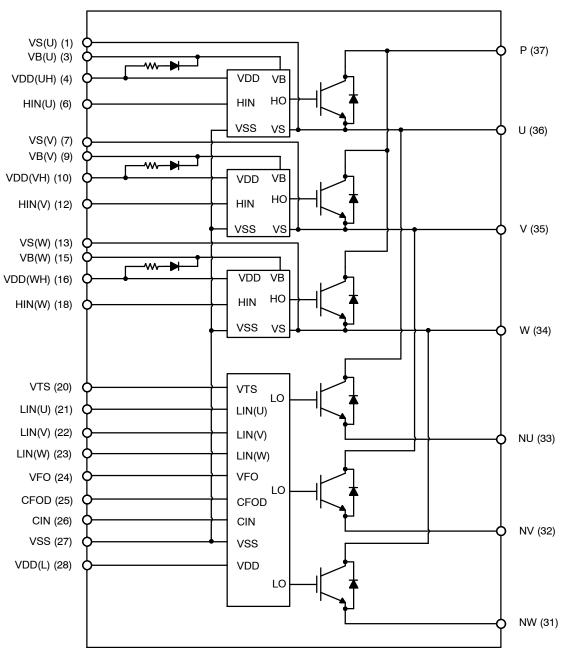
Figure 1. Pin Configuration - Top View

PIN DESCRIPTION

Pin	Name	Description
1	VS(U)	High-Side Bias Voltage Ground for U-Phase IGBT Driving
(2)		Dummy
3	VB(U)	High-Side Bias Voltage for U-Phase IGBT Driving
4	VDD(UH)	High-Side Bias Voltage for U-Phase IC
(5)	_	Dummy
6	HIN(U)	Signal Input for High-Side U-Phase
7	VS(V)	High-Side Bias Voltage Ground for V-Phase IGBT Driving
(8)	-	Dummy
9	VB(V)	High-Side Bias Voltage for V-Phase IGBT Driving
10	VDD(VH)	High-Side Bias Voltage for V-Phase IC
(11)	-	Dummy
12	HIN(V)	Signal Input for High-Side V-Phase
13	VS(W)	High-Side Bias Voltage Ground for W-Phase IGBT Driving
(14)	-	Dummy
15	VB(W)	High-Side Bias Voltage for W-Phase IGBT Driving
16	VDD(WH)	High-Side Bias Voltage for W-Phase IC
(17)	=	Dummy
18	HIN(W)	Signal Input for High-Side W-Phase
(19)	=	Dummy
20	VTS	Output for LVIC Temperature Sensing Voltage
21	LIN(U)	Signal Input for Low-Side U-Phase
22	LIN(V)	Signal Input for Low-Side V-Phase
23	LIN(W)	Signal Input for Low-Side W-Phase
24	VFO	Fault Output
25	CFOD	Capacitor for Fault Output Duration Selection
26	CIN	Input for Over Current Protection
27	VSS	Low-Side Common Supply Ground
28	VDD(L)	Low-Side Bias Voltage for IC and IGBTs Driving
(29)	-	Dummy
(30)	-	Dummy
31	NW	Negative DC-Link Input for W-Phase
32	NV	Negative DC-Link Input for V-Phase
33	NU	Negative DC-Link Input for U-Phase
34	W	Output for W-Phase
35	V	Output for V-Phase
36	U	Output for U-Phase
37	Р	Positive DC-Link Input
38	N.C	No Connection
(39)	_	Dummy

NOTE: Pins of () are the dummy for internal connection. These pins should be no connection.

INTERNAL EQUIVALENT CIRCUIT AND INPUT/OUTPUT PINS



ABSOLUTE MAXIMUM RATINGS (VDD = 15 V and T_i = 25 °C, Unless Otherwise Specified)

Symbol	Rating	Conditions	Value	Unit
INVERTER PA	RT			
VPN	Supply Voltage	Applied between P – NU, NV, NW	900	V
VPN(Surge)	Supply Voltage (Surge)	Applied between P – NU, NV, NW (Note 1)	1000	V
Vces	Vces Collector - Emitter Voltage		1200	V
VRRM	Maximum Repetitive Revers Voltage		1200	V
±lc	Each IGBT Collector Current		15	Α
±lcp	Each IGBT Collector Current (Peak)	Tc = 25 °C, Tj ≤ 150 °C, under 1 ms Pulse Width	30	Α
Pc	Corrector Dissipation	Tc = 25 °C per one chip (Note 2)	109	W
Tj	Operating Junction Temperature		−40 ~ 150	°C
CONTROL PAI	RT			
VDD	Control Supply Voltage	Applied between VDD(H), VDD(L) - VSS	20	V
VBS	High-Side Control Bias Voltage	Applied between VB(U) – VS(U), VB(V) – VS(V), VB(W) – VS(W)	20	V
VIN	Input Signal Voltage	Applied between HIN(U), HIN(V), HIN(W), LIN(U), LIN(V), LIN(W) – VSS	-0.3 ~ VDD + 0.3	V
VFO	Fault Output Supply Voltage	Applied between VFO – VSS	-0.3 ~ VDD + 0.3	V
IFO	Fault Output Current	Sink Current at VFO pin	2	mA
VCIN	Current Sensing Input Voltage	Applied between CIN – VSS	-0.3 ~ VDD + 0.3	V
TOTAL SYSTE	М			
VPN(PROT)	Self-Protection Supply Voltage Limit (Short Circuit Protection Capability)	VDD = VBS = 13.5 \sim 16.5 V, Tj = 150 °C, Non-repetitive, <2 μs	800	V
Tc	Case Operation Temperature	See Figure 1	−40 ~ 125	°C
Tstg	Storage Temperature		−40 ~ 125	°C
Viso	Isolation Voltage	60 Hz, Sinusoidal, AC 1 minute, Connection Pins to Heat Sink Plate	2500	V _{rms}

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

THERMAL CHARACTERISTICS

Symbol	Rating	Conditions	Min	Тур	Max	Unit
Rth(j-c)Q	Junction to Case Thermal	Inverter IGBT Part (per 1/6 Module)	-	-	1.15	°C/W
Rth(j-c)F	Resistance (Note 3)	Inverter FWDi Part (per 1/6 Module)	-	ı	1.80	°C/W

^{3.} For the measurement point of case temperature (Tc), please refer to Figure 1.

^{1.} Surge voltage developed by the switching operation due to the wiring inductance between P and NU, NV, NW terminal.

^{2.} Calculation value considered to design factor.

ELECTRICAL CHARACTERISTICS (VDD = 15 V and T_i = 25 °C, Unless Otherwise Specified)

Sy	mbol	Parameter	Test Conditions	<u> </u>	Min	Тур	Max	Unit
INVE	RTER PA	ART						
ļ	ces	Collector-Emitter Leakage	Tj = 25 °C, Vce = Vces		_	_	1	mA
		Current	Tj = 150 °C, Vce = Vces		_	-	10	mA
VC	E(sat)	Collector-Emitter Saturation	VDD = VBS = 15 V, lc = 10 A, Tj = 2	25 °C	-	1.50	1.90	V
		Voltage	VDD = VBS = 15 V, lc = 10 A, Tj =	150 °C	_	1.75	_	V
,	VF	FWDi Forward Voltage	VIN = 0 V, IF = 10 A, Tj = 25 °C		_	1.70	2.10	V
			VIN = 0 V, IF = 10 A, Tj = 150 °C		_	1.65	_	V
HS	ton	High Side Switching Times	VPN = 600 V, VDD = 15 V, lc = 10 A	A, Ti = 25 °C,	1.00	1.30	1.90	μS
	tc (on)		Inductive Load Switching See Figures 3, 23, 24	Inductive Load Switching		0.15	0.55	μS
	toff	1	(Note 4)		_	1.50	2.00	μS
	tc (off)	-			_	0.27	0.30	μS
	trr	1				0.26	_	μS
LS	ton	Low Side Switching Times			1.00	1.45	1.90	μS
	tc (on)				-	0.17	0.55	μS
	toff	-			_	1.65	2.00	μS
	tc (off)	-			_	0.24	0.30	μS
	trr	-				0.25	-	+ -
CON.	TROL PA	DT				0.25		μS
		1	\\DD(\\ \	VDD(UH) - VSS	1		0.20	
IQ	DDH	Quiescent VDD Supply Current	VDD(UH,VH,WH) = 15 V, HIN(U,V,W) = 0 V	VDD(VH) - VSS VDD(WH) - VSS VDD(WH) - VSS	_	_	0.30	mA
IQ	DDL		VDD(L) = 15 V, LIN(U, V, W) = 0 V	VDD(L) - VSS	-	-	2.00	mA
IPDDH		Operating VDD Supply Current	VDD(UH, VH, WH) = 15 V, f _{PWM} = 20 kHz, Duty = 50%, applied to one PWM Signal Input for High-Side	VDD(UH) - VSS VDD(VH) - VSS VDD(WH) - VSS	-	-	0.40	mA
ΙP	DDL		VDD(L) = 15 V, f _{PWM} = 20 kHz, Duty = 50%, applied to one PWM Signal Input for Low-Side	VDD(L) - VSS	-	-	5.00	mA
IC	QBS	Quiescent VBS Supply Current	VBS(U, V, W) = 15 V HIN(U, V, W) = 0 V	VB(U) - VS(U) VB(V) - VS(V) VB(W) - VS(W)	_	-	0.30	mA
IF	PBS	Operating VBS Supply Current	VDD(UH, VH, WH) = VBS = 15 V, f _{PWM} = 20 kHz, Duty = 50%, Applied to one PWM Signal Input for High-Side	VB(U) - VS(U) VB(V) - VS(V) VB(W) - VS(W)	-	-	3.50	mA
VIN	N(ON)	ON Threshold Voltage	HIN(U, V, W) - VSS, LIN(U, V, W) -	- VSS	-	-	2.60	٧
VIV	N(OF)	OFF Threshold Voltage			0.80	-	-	V
VC	IN(ref)	Over Current Trip Level	VDD = 15 V	CIN-VSS	0.46	0.48	0.50	٧
U٧	/DDD	Supply Circuit	Detection Level	•	10.30	_	12.50	٧
U٧	/DDR	Under-Voltage Protection	Reset Level		10.80	_	13.00	V
U۱	/BSD	1	Detection Level		10.00	_	12.00	V
U۱	/BSR	1	Reset Level		10.50	_	12.50	V
	/TS	Voltage Output for LVIC Temperature Sensing Unit	VTS-VSS = 5.1 k Ω , Temp. = 25 °C (Note 5)		1.145	1.25	1.38	٧

ELECTRICAL CHARACTERISTICS (VDD = 15 V and T_i = 25 °C, Unless Otherwise Specified) (continued)

CLECTRICAL CHARACTERISTICS (VDD = 15 V and T _j = 25 °C, Offices Office wise Specifical) (continued)								
Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit		
CONTROL PART								
VFOH	Fault Output Voltage	VDD(L) = 0 V, CIN = 0 V, VFO Circuit: 10 kΩ to 5 V Pull-up	4.90	-	_	V		
VFOL]	VDD(L) = 0 V, CIN = 1 V, VFO Circuit: 10 kΩ to 5 V Pull-up	-	_	0.95	٧		
tFOD	Fault-Output Pulse Width	CFOD = 22 nF (Note 6)	1.60	2.40	-	ms		
BOOTSTRAF	PART							
VF	Bootstrap Diode Forward Current	If = 0.1 A (See Figure 6)	2.10	2.50	2.90	٧		
RBOOT	Built-in Limiting Resistance		12.50	15.50	18.50	Ω		

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

Performance guaranteed over the indicated operating temperature range by design and/or characterization tested at Tj = Ta = 25 °C. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible. Values based on design and/or characterization.

- 4. ton and toff include the propagation delay of the internal drive IC. tc(on) and tc(off) are the switching times of IGBT under the given gate-driving condition internally. For the detailed information, please see *Figure 3*.
- 5. TLVIC is the temperature of LVIC itself. VTS is only for sensing temperature of LVIC and cannot shutdown IGBTs automatically. The relationship between VTS voltage output and LVIC temperature is described in *Figure 4*. It is recommended to add 5.1 kΩ pull down resistor between VTS and VSS (Signal Ground) as described in *Figure 5* for linear output characteristics at low temperature. Refer to the application note for usage of VTS.
- 6. The fault-out pulse width tFOD depends on the capacitance value of CFOD according to the following approximate equation: tFOD = 0.11 × 10⁶ × CFOD [s].

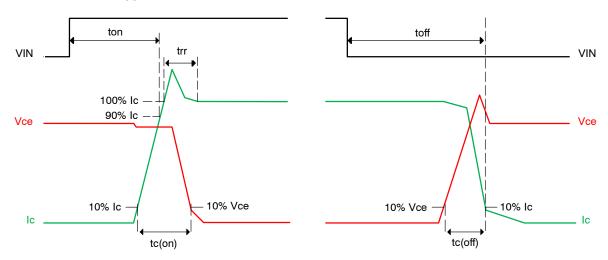


Figure 3. Switching Time Definitions

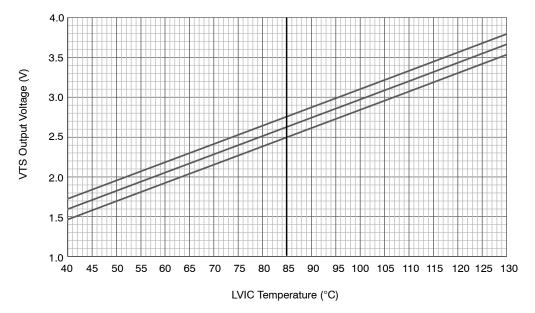


Figure 4. Temperature of LVIC vs. VTS Characteristics

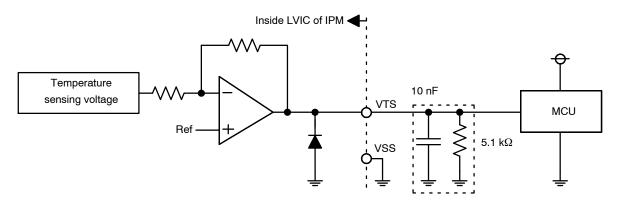


Figure 5. Internal Block Diagram and Interface Circuit of VTS

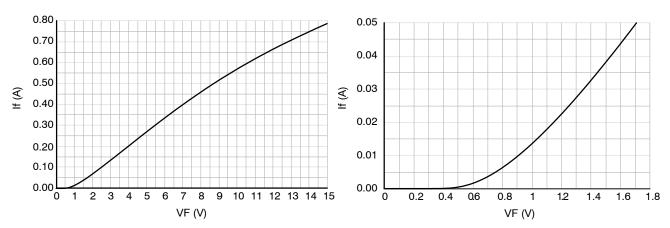


Figure 6. Characteristics of Bootstrap Diode/Resistor (Right Figure is Enlarged Figure)

RECOMMENDED OPERATING CONDITIONS

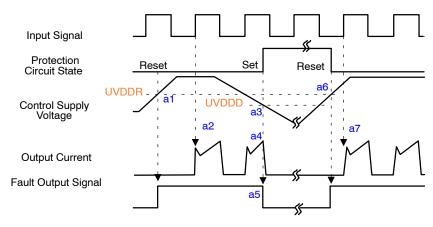
				Value			
Symbol	Description	Conditions		Min	Тур	Max	Unit
VPN	Supply Voltage	Applied between P - NU, NV, NW		_	600	800	٧
VDD	Control Supply Voltage	Applied between VDD(H) - VSS, VI	DD(L) – VSS	13.5	15	16.5	V
VBS	High-Side Bias Voltage	Applied between VB(U) – VS(U), VB(W) – VS(W)	3(V) – VS(V),	13.0	15	18.5	V
dVDD / dt, dVBS / dt	Control Supply Variation			-1	-	1	V/μs
Tdead	Blanking Time for Preventing Arm – Short	For Each Input Signal		1	-	_	μs
fPWM	PWM Input Signal	-40 °C ≤ Tc ≤ 125 °C, -40°C ≤ Tj ≤	150 °C	1	_	20	kHz
lo	Allowable r.m.s. Current	VPN = 600 V, VDD = VBS = 15 V,	fPWM = 5 kHz	_	_	15.3	Arms
		P.F. = 0.8, Tc ≤ 125 °C, Tj ≤ 150 °C (Note 7)	fPWM = 15 kHz	-	-	9.5	Arms
PWIN(ON)	Minimum Input Pulse Width	VDD = VBS = 15 V, Wiring Inductance between NU, V, W and DC Link N < 10 nH (Note 8)		1.0	_	-	μs
PWIN (OFF)				2.0	-	_	
Package Mo	unting Torque	M3 Type Screw		0.6	0.7	0.9	Nm

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

Flatness tolerance of the heatsink should be within –50 μm to +100 μm .

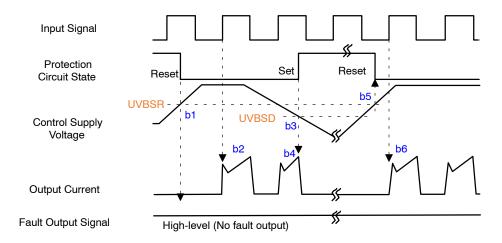
- 7. Allowable output current value is the reference data for the safe operation of this product. This may be different from the actual application and operating condition.
- 8. Product might not make response if input pulse width is less than the recommended value.

TIME CHARTS OF PROTECTIVE FUNCTION



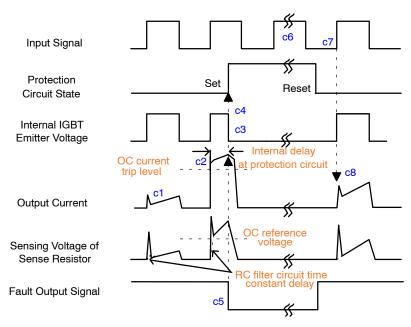
- a1: Control supply voltage rises: After the voltage rises UVDDR, the circuits start to operate when next input is applied.
- a2: Normal operation: IGBT ON and carrying current.
- a3: Under voltage detection (UVDDD).
- a4: IGBT OFF in spite of control input condition.
- a5: Fault output operation starts with a fixed pulse width. a6: Under voltage reset (UVDDR).
- a7: Normal operation: IGBT ON and carrying current by triggering next signal from LOW to HIGH.

Figure 7. Under-Voltage Protection (Low-Side)



- b1: Control supply voltage rises: After the voltage reaches UVBSR, the circuits start to operate when next input is applied.
- b2 : Normal operation: IGBT ON and carrying current.
- b3: Under voltage detection (UVBSD).
- b4: IGBT OFF in spite of control input condition, but there is no fault output signal.
- b5: Under voltage reset (UVBSR).
- b6: Normal operation: IGBT ON and carrying current by triggering next signal from LOW to HIGH.

Figure 8. Under-Voltage Protection (High-Side)

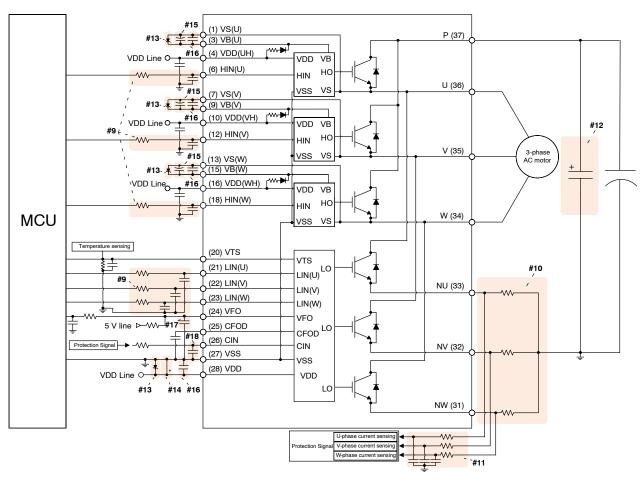


(with the external sense resistance and RC filter connection)

- c1: Normal operation: IGBT ON and carrying current.
- c2 : Over current detection (OC trigger).
- c3: All low-side IGBT's gate are hard interrupted.
- c4 : All low-side IGBTs turn OFF.
- c5 : Fault output operation starts with a fixed pulse width.
- c6: Input HIGH: IGBT ON state, but during the active period of fault output the IGBT doesn't turn ON.
- c7 : Fault output operation finishes, but IGBT doesn't turn on until triggering next signal from LOW to HIGH.
- c8: Normal operation: IGBT ON and carrying current.

Figure 9. Over Current Protection (Low-Side Operation only)

TYPICAL APPLICATION CIRCUIT

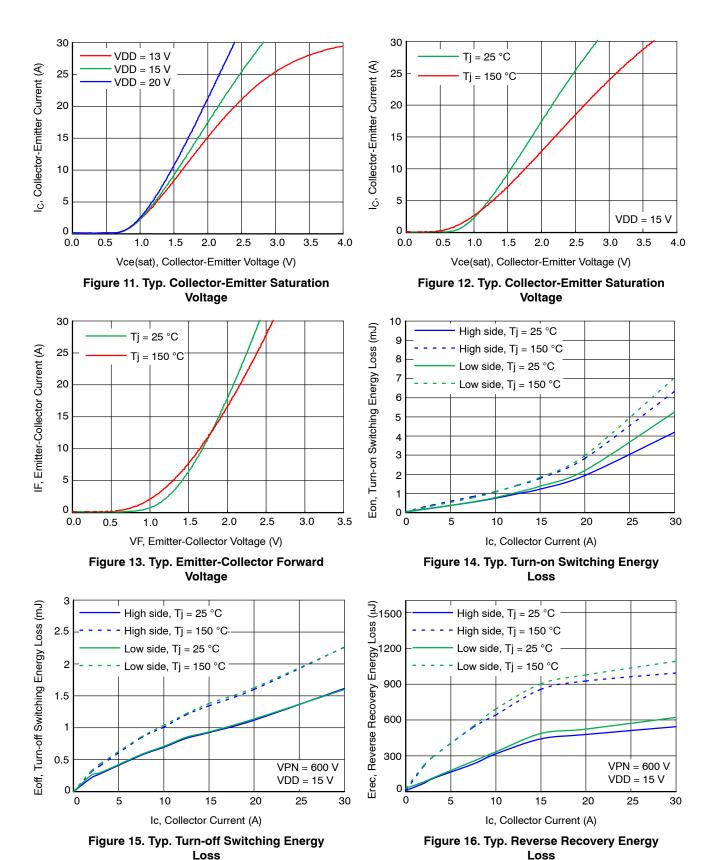


To avoid malfunction, the wiring of each input should be as short as possible (less than 2–3 cm). Each capacitor should be mounted as close to the pins of the product as possible. VFO output is open-drain type. This signal line should be pulled up to the positive side of the MCU or control power supply with a resistor that makes IFO up to 1mA. Please refer to Figure 5.

- 9. Input signal is active-HIGH type. There is a 5 k Ω resistor inside the IC to pull-down each input signal line to GND. RC coupling circuits should be adopted for the prevention of input signal oscillation. RC time constant should be selected in the range 50 ~ 150 ns. (Recommended R = 100 Ω , C = 1 nF)
- 10. Each wiring pattern inductance should be minimized (Recommend less than 10 nH). Use the shunt resistor of surface mounted (SMD) type to reduce wiring inductance. To prevent malfunction, wiring should be connected to the terminal of the shunt resistor as close as possible.
- 11. In the short-circuit protection circuit, please select the RC time constant in the range 1.5 ~ 2 μs. Do enough evaluation on the real system because short-circuit protection time may vary wiring pattern layout and value of the RC time constant.
- 12. To prevent surge destruction, the wiring between the snubber capacitor and the P & GND pins should be as short as possible. The use of a high-frequency non-inductive capacitor of around $0.1 \sim 0.22~\mu F$ between the P & GND pins is recommended.
- 13. The zener diode or transient voltage suppressor should be adopted for the protection of ICs from the surge destruction between each pair of control supply terminals (Recommended zener diode is 22 V / 1 W, which has the lower zener impedance characteristic than about 15 Ω).
- 14. VDD electrolytic capacitor is recommended around 7 times larger than VBS electrolytic bootstrap capacitor.
- 15. Please choose the VBS electrolytic bootstrap capacitor with good temperature characteristic.
- 16.0.1 ~ 0.2 μF R-category ceramic capacitors with good temperature and frequency characteristics is recommended.
- 17. Fault out pulse width can be adjusted by capacitor connected to the CFOD terminal.
- 18. To prevent protection function errors, CIN capacitor should be placed as close to CIN and VSS pins as possible.

Figure 10. Typical Application Circuit

TYPICAL CHARACTERISTICS



TYPICAL CHARACTERISTICS (CONTINUED)

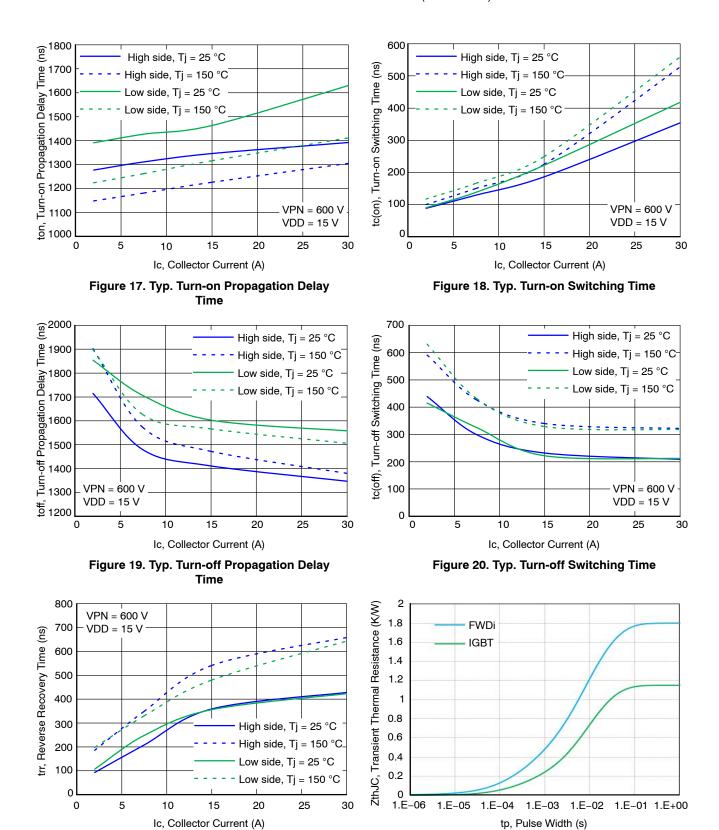


Figure 21. Typ. Reverse Recovery Time

Figure 22. Transient Thermal Resistance

TURN-ON/OFF SWITCHING WAVEFORM

SWITCHING CONDITION: VDC = 600 V, VDD = 15 V, TJ = 25 $^{\circ}$ C, IC = 10 A.

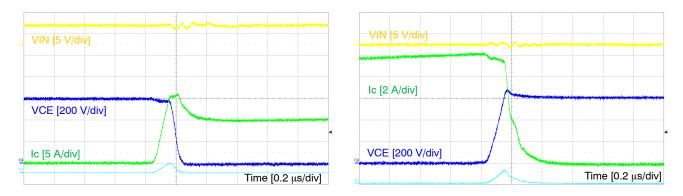


Figure 23. Turn-on Switching Waveform

Figure 24. Turn-off Switching Waveform

REVISION HISTORY

Revision	Description of Changes	Date
4	Updated MIN limit spec of VTS item in the Electrical Characteristics Table (CONTROL PART SECTION) on page 6.	7/1/2025

This document has undergone updates prior to the inclusion of this revision history table. The changes tracked here only reflect updates made on the noted approval dates.







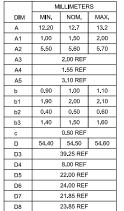
DIP39, 54.50x31.00x5.60, 1.78P EP-2 CASE MODGX **ISSUE B**

DATE 05 MAY 2025

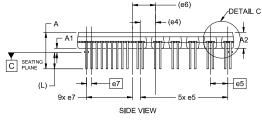
NOTES

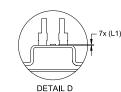
- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 2009. CONTROLLING DIMENSION: MILLIMETERS
- DIMENSION b and c APPLY TO THE PLATED LEADS AND ARE MEASURED BETWEEN 1.00 AND 2.00 FROM THE LEAD TIP POSITION OF THE LEAD IS DETERMINED AT THE BASE OF THE 3.
- LEAD WHERE IT EXITS THE PACKAGE BODY
- AREA FOR 2D BAR CODE
- SHORTENED/CUT PINS ARE 2,5,8,11,14,17,19,29, 30 AND 39
- E SIDE FLASH

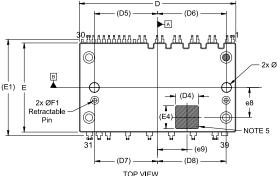
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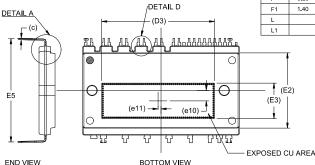


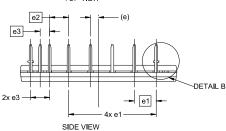
	MILLIMETERS			
DIM	MIN.	MAX.		
E	30.90	31.00	31,10	
E1		33.50 REF		
E2		26.14 REF		
E3		12.35 REF		
E4		8.00 REF		
E5	35.40	35.90	36.40	
е		2.81 REF		
e1		7.62 BSC		
e2	6.60 BSC			
e3		3.30 BSC		
e4		5.35 REF		
e5		6.10 BSC		
e6		8.02 REF		
e7		1.78 BSC		
e8		10.35 REF		
e9		10.25 REF		
e10		3.60 REF		
e11		1.00 REF		
e12		0.89 BSC		
F	3.20	3.30	3.40	
F1	1.40	1.50	1.60	
L	5.60 REF			
L1		0.10 REF		

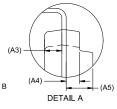


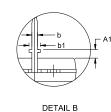


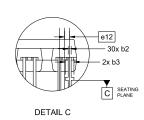




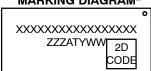








GENERIC MARKING DIAGRAM*



XXXXX = Specific Device Code

= Assembly Lot Code 777

ΑT = Assembly & Test Location

= Year WW = Work Week *This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot " • ", may or may not be present. Some products may not follow the Generic Marking.

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