SiC JFET Division

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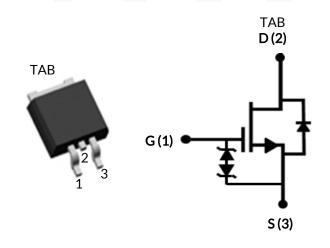
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DATASHEET

UF3C065030B3



Part Number	Package	Marking
UF3C065030B3	D ² PAK-3L	UF3C065030B3



Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, D2PAK-3L, 650 V, 27 mohm

Rev. D, January 2025

Description

This SiC FET device is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the D²PAK-3L package, this device exhibits ultralow gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads when used with recommended RC-snubbers, and any application requiring standard gate drive.

Features

- Typical on-resistance R_{DS(on),typ} of 27mΩ
- Maximum operating temperature of 175°C
- Excellent reverse recovery
- Low gate charge
- Low intrinsic capacitance
- ESD protected, HBM class 2

• Very low switching losses (required RC-snubber loss negligible under typical operating conditions)

AECQ Qualified

Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating





Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V _{DS}		650	V
Gate-source voltage	V _{GS}	DC	-25 to +25	V
Continuous drain current ¹	1	T _C = 25°C	65	А
Continuous drain current	ID	T _C = 100°C	47	А
Pulsed drain current ²	I _{DM}	T _C = 25°C	230	А
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =4A	120	mJ
Power dissipation	P _{tot}	T _C = 25°C	242	W
Maximum junction temperature	T _{J,max}		175	°C
Operating and storage temperature	TJ, TSTG		-55 to 175	°C
Reflow soldering temperature	T_{solder}	reflow MSL 1	245	°C

1. Limited by $T_{J,max}$

2. Pulse width t_{p} limited by $T_{J,\text{max}}$

3. Starting $T_J = 25^{\circ}C$

Thermal Characteristics

Darameter	Symbol	Test Conditions		Units		
Parameter			Min	Тур	Max	Units
Thermal resistance, junction-to-case	$R_{\theta JC}$			0.48	0.62	°C/W

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Electrical Characteristics (T_J = +25°C unless otherwise specified)

Typical Performance - Static

Parameter	Cumpbed.	Test Conditions		Linite			
	Symbol		Min	Тур	Max	Units	
Drain-source breakdown voltage	BV _{DS}	V _{GS} =0V, I _D =1mA	650			V	
Total duain lookaga guurant		V _{DS} =650V, V _{GS} =0V, T _J =25°C		6	150		
Total drain leakage current	I _{DSS}	V _{DS} =650V, V _{GS} =0V, T _J =175°C	30			μA	
Total gate leakage current	I _{GSS}	V _{DS} =0V, T _J =25°C, V _{GS} =-20V / +20V		6	±20	μA	
Drain-source on-resistance	R _{DS(on)}	V _{GS} =12V, I _D =40A, T _J =25°C		27	35		
		V _{GS} =12V, I _D =40A, T _J =125°C		35		mΩ	
		V _{GS} =12V, I _D =40A, T _J =175°C		43			
Gate threshold voltage	V _{G(th)}	V_{DS} =5V, I_{D} =10mA	4	5	6	V	
Gate resistance	R _G	f=1MHz, open drain		4.5		Ω	

Typical Performance - Reverse Diode

Parameter	Symbol	Test Conditions		- Units		
		Test Conditions	Min	Тур	Max	Units
Diode continuous forward current ¹	ls	T _C =25°C			65	А
Diode pulse current ²	I _{S,pulse}	T _C =25°C			230	А
Forward voltage	V _{FSD}	V _{GS} =0V, I _S =20A, T _J =25°C		1.3	1.4	V
	• FSD	V _{GS} =0V, I _S =20A, T _J =175°C		1.35		• • •
Reverse recovery charge	Q _{rr}	V _R =400V, I _S =40A, V _{GS} =-5V, R _{G_EXT} =22Ω		211		nC
Reverse recovery time	t _{rr}	di/dt=1500A/µs, T_=25°C		34		ns
Reverse recovery charge	Q _{rr}	V _R =400V, I _S =40A, V _{GS} =-5V, R _{G_EXT} =22Ω		188		nC
Reverse recovery time	t _{rr}	di/dt=1500A/µs, T_=150°C		32		ns





Typical Performance - Dynamic

Deven	Course la sel	Test Conditions		L Institus			
Parameter	Symbol	Test Conditions —	Min	Тур	Max	Units	
Input capacitance	C _{iss}	V _{DS} =100V, V _{GS} =0V		1500			
Output capacitance	C _{oss}	$v_{DS} = 100 v, v_{GS} = 0 v$ = f=100kHz		293		pF	
Reverse transfer capacitance	C _{rss}			2			
Effective output capacitance, energy related	$C_{oss(er)}$	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		215		pF	
Effective output capacitance, time related	C _{oss(tr)}	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		480		pF	
C _{OSS} stored energy	E _{oss}	V _{DS} =400V, V _{GS} =0V		17.5		μJ	
Total gate charge	Q _G	– V _{DS} =400V, I _D =40A, –		51			
Gate-drain charge	Q_{GD}	$V_{\rm DS} = -5V \text{ to } 15V$		11		nC	
Gate-source charge	Q_{GS}	VGS - 5V (015V		19			
Turn-on delay time	t _{d(on)}			34			
Rise time	t _r	V _{DS} =400V, I _D =40A, Gate		16		- ns	
Turn-off delay time	$t_{d(off)}$	Driver =-5V to +15V,		56			
Fall time	t _f	Turn-on $R_{G,EXT}$ =1.8 Ω ,		15			
Turn-on energy including R _s energy ⁴	E _{ON}	$- \text{Turn-off } R_{G,EXT} = 22\Omega - $ Inductive Load,		392		μ	
Turn-off energy including R_s energy ⁴	E _{OFF}	FWD: same device with		113			
Total switching energy including R _s energy ⁴	E _{total}	V_{GS} = -5V and R_{G} = 22 Ω , RC snubber: R_{S} =5 Ω and		505			
Snubber R _s energy during turn-on	E _{RS_ON}	С _S =330pF, Т _J =25°С		5.3			
Snubber R _s energy during turn-off	E _{RS_OFF}			7.9			
Turn-on delay time	t _{d(on)}			32			
Rise time	t _r	V _{DS} =400V, I _D =40A, Gate		16			
Turn-off delay time	t _{d(off)}	Driver =-5V to +15V,		57		– ns	
Fall time	t _f	Turn-on $R_{G,EXT}$ =1.8 Ω ,		16		1	
Turn-on energy including R _s energy ⁴	E _{ON}	- Turn-off $R_{G,EXT}=22\Omega$ -		370			
Turn-off energy including R _s energy ⁴	E _{OFF}	Inductive Load, FWD: same device with $V_{GS} = -5V$ and $R_G = 22\Omega$, RC snubber: $R_S=5\Omega$ and		118		μJ	
Total switching energy including R _s energy ⁴	E _{TOTAL}			488			
Snubber R _s energy during turn-on	E _{RS_ON}	C _S =330pF, T _J =150°C		4.6			
Snubber R _s energy during turn-off	E _{RS_OFF}			8.2			

4. The switching performance are evaluated with a RC snubber circuit as shown in Figure 24.

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Typical Performance Diagrams

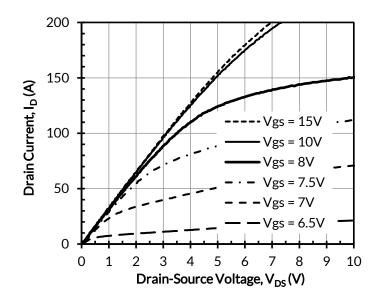


Figure 1. Typical output characteristics at $T_{\rm J}$ = - 55°C, tp < 250 μs

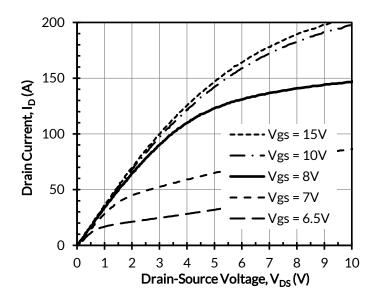


Figure 2. Typical output characteristics at T_J = 25°C, tp < 250 μ s

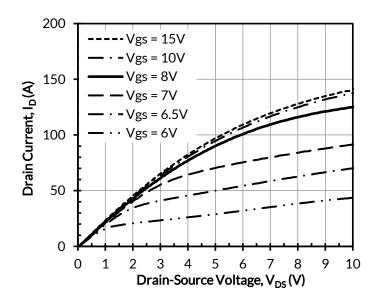


Figure 3. Typical output characteristics at T $_{\rm J}$ = 175°C, tp < 250 μs

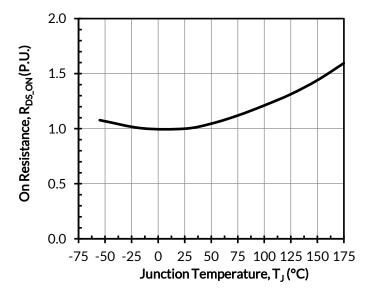


Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_D = 40A

QOULO



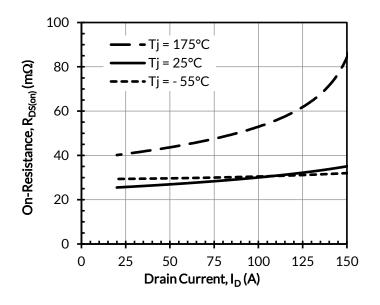


Figure 5. Typical drain-source on-resistances at V_{GS} = 12V

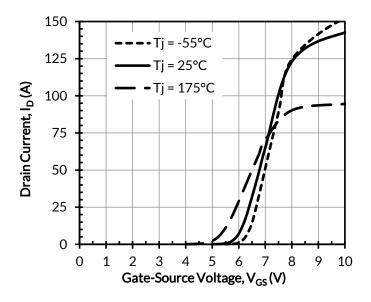


Figure 6. Typical transfer characteristics at V_{DS} = 5V

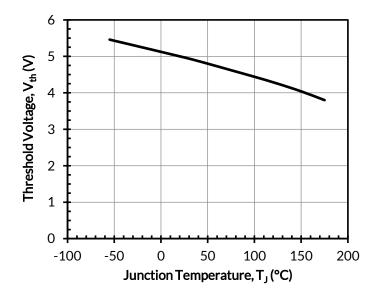


Figure 7. Threshold voltage vs. junction temperature at V_{DS} = 5V and I_{D} = 10mA

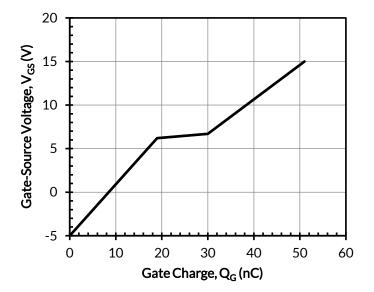
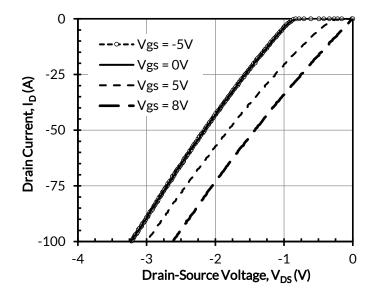


Figure 8. Typical gate charge at V_{DS} = 400V and I_{D} = 40A

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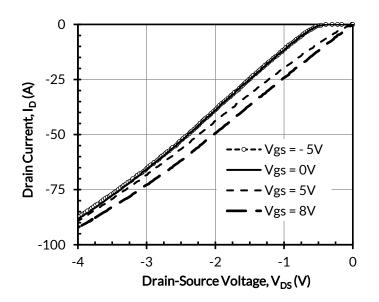


Figure 11. 3rd quadrant characteristics at T_J = 175°C

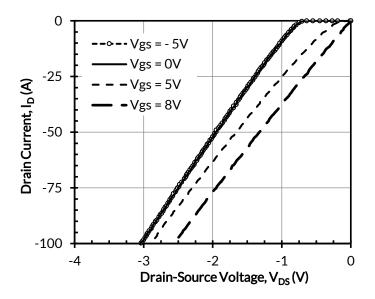


Figure 10. 3rd quadrant characteristics at T_J = 25°C

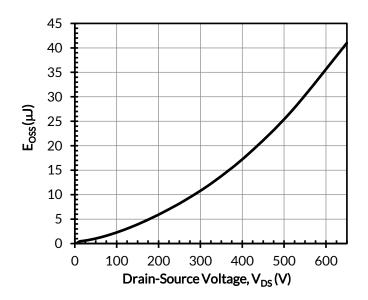


Figure 12. Typical stored energy in C_{OSS} at V_{GS} = 0V



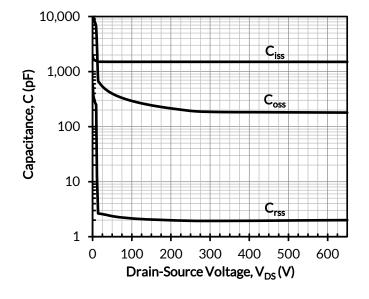


Figure 13. Typical capacitances at f = 100kHz and $V_{\rm GS}$ = 0V

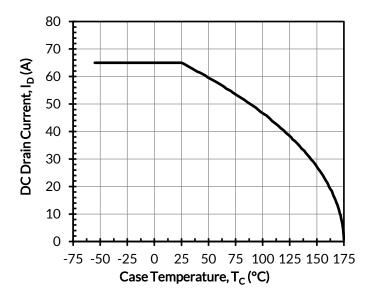


Figure 14. DC drain current derating

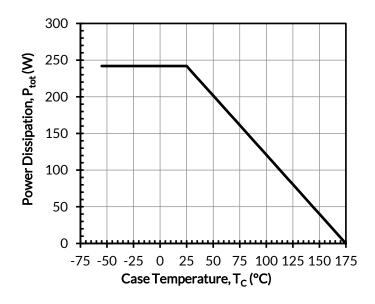


Figure 15. Total power dissipation

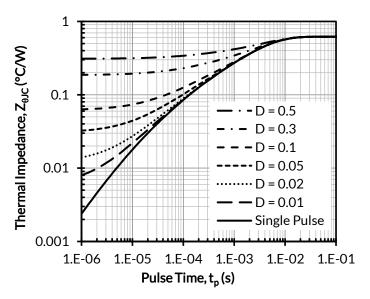


Figure 16. Maximum transient thermal impedance

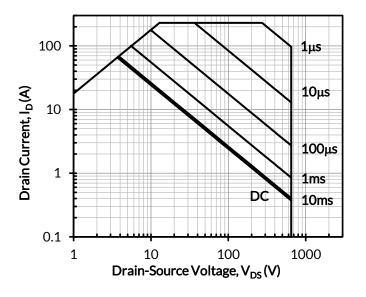
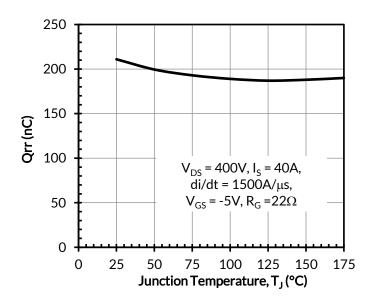


Figure 17. Safe operation area at T_{C} = 25°C, D = 0, Parameter $t_{\rm p}$



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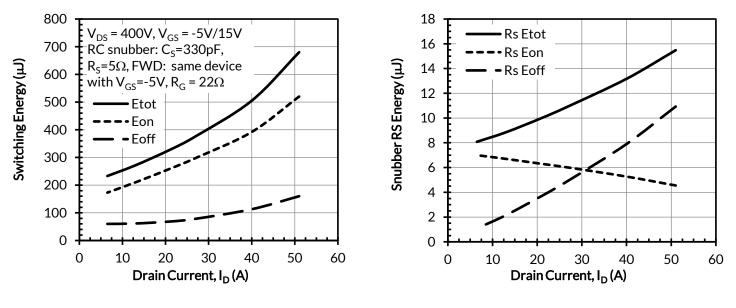
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Figure 18. Reverse recovery charge Qrr vs. junction temperture



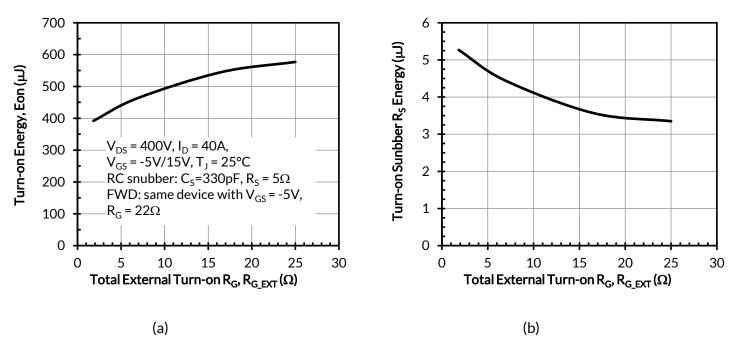
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(a)

(b)

Figure 19. Clamped inductive switching energy (a) and RC snubber energy loss (b) vs. drain current at $T_J = 25^{\circ}$ C, turn-on $R_{G_{EXT}} = 1.8\Omega$, and turn-off $R_{G_{EXT}} = 22\Omega$



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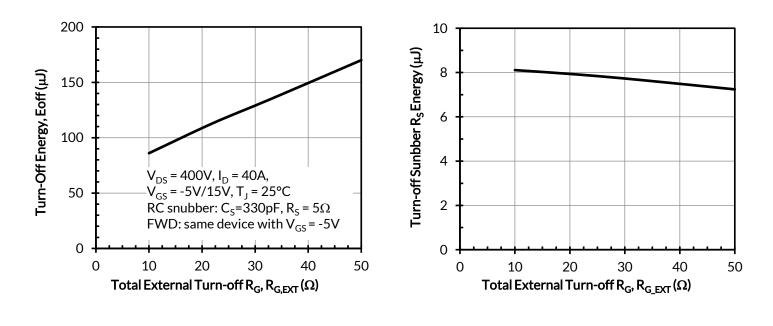
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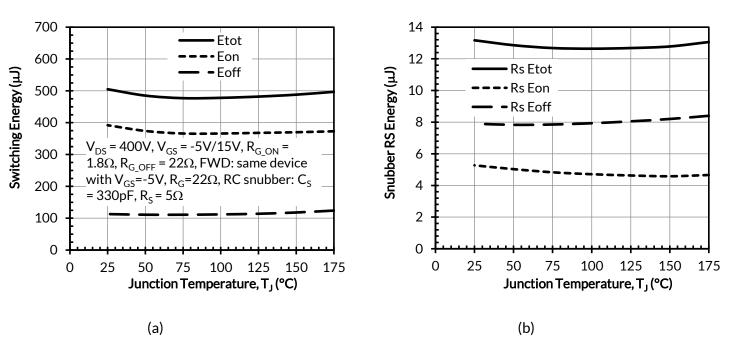
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Figure 20. Clamped inductive switching turn-on energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of total external turn-on gate resistor $R_{G_{EXT}}$



(a) (b) Figure 21. Clamped inductive switching turn-off energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of total external turn-off gate resistor $R_{G_{EXT}}$



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Figure 22. Clamped inductive switching energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of junction temperature at $I_D = 40A$

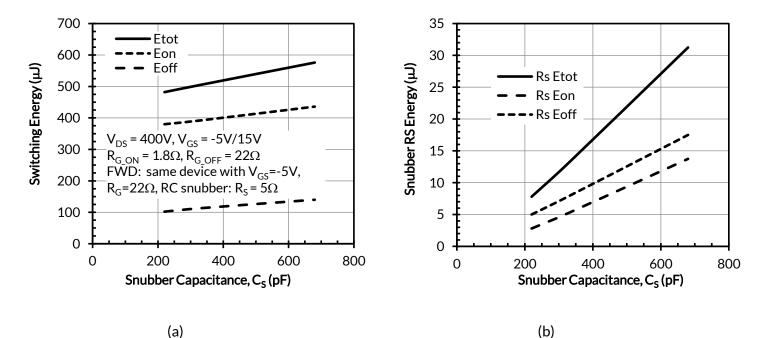


Figure 23. Clamped inductive switching energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of snubber capacitance at $I_D = 40A$ and $T_J = 25^{\circ}C$

QONO



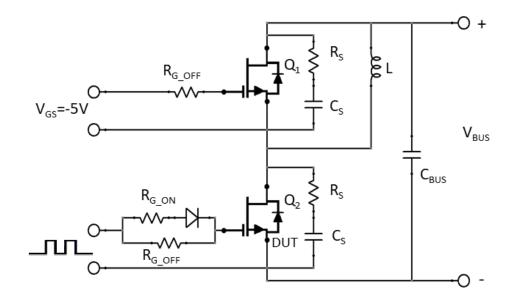


Figure 24. Clamped inductive load switching test circuit An RC snubber ($R_s = 5\Omega$ and $C_s = 330$ pF) is required to improve the turn-off waveforms.

Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com





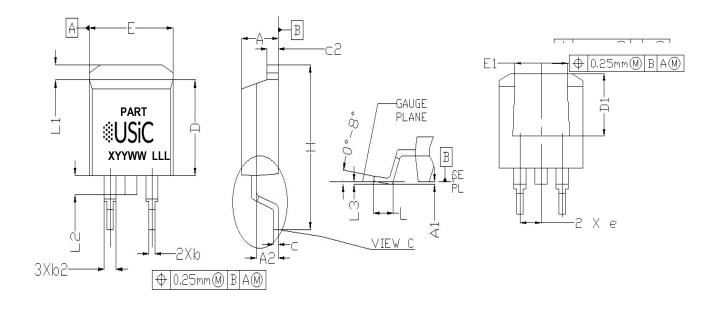
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TO263 (D2PAK)-3L PACKAGE OUTLINE, PART MARKING AND TAPE AND REEL SPECIFICATIONS

PACKAGE OUTLINE

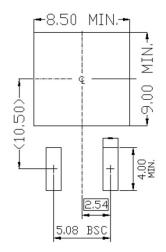


SYM	INC	HES	MILLIN	NETERS	
	MIN	ΜΑΧ	MIN	МАХ	
А	0.160	0.190	4.064	4.826	
A1	0.000	0.010	0.00	0.254	
A2	0.087	0.114	2.20	2.8956	
b	0.020	0.039	0.508	0.9906	
b2	0.045	0.07	1.143	1.778	
с	0.015	0.029	0.381	0.7366	
c2	0.045	0.065	1.143	1.651	
D	0.330	0.380	8.382	9.652	
D1	0.270	0.330	6.858	8.37	
е	0.100 BSC		2.54	BSC	
E	0.380	0.420	9.652	10.668	
E1	0.245	0.330	6.223	8.37	
Н	0.575	0.625	14.605	15.875	
L	0.070	0.110	1.778	2.794	
L1	0.040	0.066	1.02	1.6764	
L2	0.050	0.07	1.27	1.778	
L3	0.010) BSC	0.25 BSC		



TO263 (D2PAK)-3L PACKAGE OUTLINE, PART MARKING AND TAPE AND REEL SPECIFICATIONS

PCB LAND PATTERN



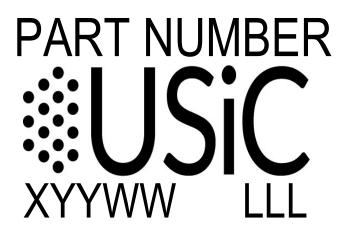
Notes:

- 1. PACKAGE BODY SIDES EXCLUDE MOLD FLASH AND GATE BURRS.
- 2. TOLERANCE 0.10MILLIMETERS UNLESS OTHERWISE SPECIFIED.
- 3. DIMENSION L IS MEASURED IN GAUGE LINE.
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PART MARKING



PART NUMBER = REFER TO DS_PN DECODER FOR DETAILS

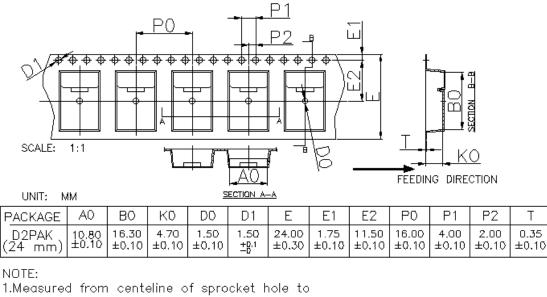
X = ASSEMBLY SITE YY = YEAR WW = WORK WEEK LLL = LOT ID

PACKING TYPE

ANTI-STATIC TAPE & REEL (T&R)

QUANTITY / REEL : 800 UNITS

CARRIER TAPE DRAWING



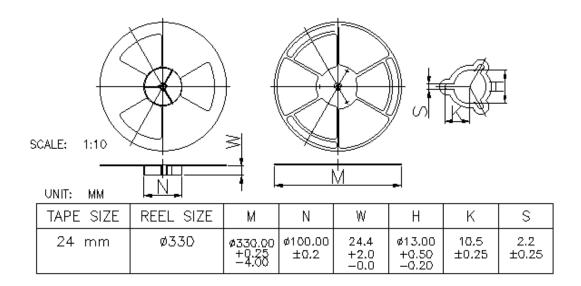
centreline of pocket.

2.Cumulative tolerance of 10 sprocket holes is ± 0.20 .

3.Camber not to exceed 2mm in 200mm



REEL DRAWING



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