

# **SiC JFET Division**

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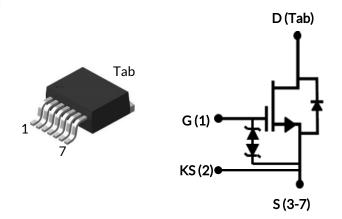






## DATASHEET

# F3SC120040B7S



Part Number	Package	Marking
UF3SC120040B7S	D <sup>2</sup> PAK-7L	UF3SC120040B7S







# Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, D2PAK-7L, 1200 V, 35 mohm

Rev. C, January 2025

### Description

This SiC FET device is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the D<sup>2</sup>PAK-7L package, this device exhibits ultralow gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads, and any application requiring standard gate drive.

#### **Features**

- On-resistance R<sub>DS(on)</sub>: 35mΩ (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q<sub>rr</sub> = 358nC
- Low body diode V<sub>FSD</sub>: 1.5V
- ◆ Low gate charge: Q<sub>G</sub> = 43nC
- Threshold voltage V<sub>G(th)</sub>: 5V (typ) allowing 0 to 15V drive
- Package creepage and clearance distance > 6.1mm
- Kelvin source pin for optimized switching performance
- ESD protected, HBM class 2

## Typical applications

Any controlled environment such as

- ◆ Telecom and Server Power
- Industrial power supplies
- Power factor correction modules
- Motor drives
- Induction heating













# **Maximum Ratings**

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	$V_{DS}$		1200	V
Gate-source voltage	$V_{GS}$	DC	-25 to +25	V
Continuous drain current <sup>1</sup>	1	T <sub>C</sub> = 25°C	47	Α
Continuous drain current	I <sub>D</sub>	T <sub>C</sub> = 100°C	34	Α
Pulsed drain current <sup>2</sup>	I <sub>DM</sub>	T <sub>C</sub> = 25°C	175	Α
Single pulsed avalanche energy <sup>3</sup>	E <sub>AS</sub>	L=15mH, I <sub>AS</sub> =4.2A	132.3	mJ
Power dissipation	P <sub>tot</sub>	T <sub>C</sub> = 25°C	214	W
Maximum junction temperature	$T_{J,max}$		175	°C
Operating and storage temperature	$T_J,T_STG$		-55 to 175	°C
Reflow soldering temperature	$T_{solder}$	reflow MSL 3	245	°C

- 1. Limited by  $T_{J,max}$
- 2. Pulse width  $t_p$  limited by  $T_{J,max}$
- 3. Starting  $T_J = 25^{\circ}C$

# **Thermal Characteristics**

Darameter			Value		Units	
Parameter	Symbol Test Conditions	Min	Тур	Max	Units	
Thermal resistance, junction-to-case	$R_{ heta$ JC			0.54	0.7	°C/W













# Electrical Characteristics ( $T_J = +25$ °C unless otherwise specified)

# **Typical Performance - Static**

Parameter	Symbol	Test Conditions	Value			Units	
rai afficter	Syllibol	Symbol		Тур	Max	Offics	
Drain-source breakdown voltage	BV <sub>DS</sub>	$V_{GS}$ =0V, $I_D$ =1mA	1200			V	
Tatal dualin lackage august		$V_{DS}$ =1200V, $V_{GS}$ =0V, $T_{J}$ =25°C		8	150		
Total drain leakage current	I <sub>DSS</sub>	V <sub>DS</sub> =1200V, V <sub>GS</sub> =0V, T <sub>J</sub> =175°C		35		μΑ	
Total gate leakage current	I <sub>GSS</sub>	V <sub>DS</sub> =0V, T <sub>J</sub> =25°C, V <sub>GS</sub> =-20V / +20V		6	±20	μА	
Drain-source on-resistance	R <sub>DS(on)</sub>	$V_{GS}$ =12V, $I_D$ =35A, $T_J$ =25°C		35	45	mΩ	
Drain source on resistance		V <sub>GS</sub> =12V, I <sub>D</sub> =35A, T <sub>J</sub> =175°C		73		11122	
Gate threshold voltage	$V_{G(th)}$	$V_{DS}$ =5V, $I_D$ =10mA	4	5	6	V	
Gate resistance	$R_{G}$	f=1MHz, open drain		4.5		Ω	

# Typical Performance - Reverse Diode

Deversation	Cymahal Tae	Toot Conditions	Value				
Parameter	Symbol	Test Conditions	Min	Тур	Max	Units	
Diode continuous forward current <sup>1</sup>	I <sub>S</sub>	T <sub>C</sub> =25°C			47	Α	
Diode pulse current <sup>2</sup>	$I_{S,pulse}$	T <sub>C</sub> =25°C			175	Α	
Forward voltage	$V_{FSD}$	V <sub>GS</sub> =0V, I <sub>S</sub> =20A, T <sub>J</sub> =25°C		1.5	2	V	
1 of ward voltage	▼ FSD	V <sub>GS</sub> =0V, I <sub>S</sub> =20A, T <sub>J</sub> =175°C		1.95		•	
Reverse recovery charge	$Q_{rr}$	$V_R$ =800V, $I_S$ =40A, $V_{GS}$ =-5V, $R_{G_EXT}$ =10 $\Omega$		358		nC	
Reverse recovery time	t <sub>rr</sub>	di/dt=2400A/μs, T <sub>J</sub> =25°C		25		ns	
Reverse recovery charge	$Q_{rr}$	V <sub>R</sub> =800V, I <sub>S</sub> =40A, V <sub>GS</sub> =-5V, R <sub>G_EXT</sub> =10Ω		259		nC	
Reverse recovery time	t <sub>rr</sub>	di/dt=2400A/μs, Τ <sub>J</sub> =150°C		22		ns	













# Typical Performance - Dynamic

Danamatan	Symbol Test Conditions —	Value			Linita	
Parameter	, ,		Min	Тур	Max	Units
Input capacitance	C <sub>iss</sub>	V <sub>DS</sub> =100V, V <sub>GS</sub> =0V		1500		
Output capacitance	C <sub>oss</sub>	f=100kHz		210		pF
Reverse transfer capacitance	$C_{rss}$	1-100KH2		1.7		
Effective output capacitance, energy related	C <sub>oss(er)</sub>	$V_{DS}$ =0V to 800V, $V_{GS}$ =0V		112		pF
Effective output capacitance, time related	C <sub>oss(tr)</sub>	$V_{DS}$ =0V to 800V, $V_{GS}$ =0V		280		pF
C <sub>OSS</sub> stored energy	E <sub>oss</sub>	$V_{DS}$ =800V, $V_{GS}$ =0V		35.6		μЈ
Total gate charge	$Q_{G}$	- V <sub>DS</sub> =800V, I <sub>D</sub> =35A, -		43		
Gate-drain charge	$Q_{GD}$	$V_{DS} = -5V \text{ to } 12V$		11		nC
Gate-source charge	$Q_{GS}$	VGS 3 V to 12 V		19		
Turn-on delay time	t <sub>d(on)</sub>	$V_{DS}$ =800V, $I_D$ =35A, Gate Driver =-5V to +12V, Turn-on $R_{G,EXT}$ =8.5 $\Omega$ ,		40		ns
Rise time	t <sub>r</sub>			13		
Turn-off delay time	t <sub>d(off)</sub>			47		
Fall time	t <sub>f</sub>	Turn-off $R_{G,EXT}$ =22 $\Omega$ Inductive Load.		8		
Turn-on energy	E <sub>ON</sub>	FWD: same device with		731		
Turn-off energy	E <sub>OFF</sub>	$V_{GS} = -5V, R_G = 22\Omega,$		130		μJ
Total switching energy	E <sub>TOTAL</sub>	T <sub>J</sub> =25°C		861		
Turn-on delay time	t <sub>d(on)</sub>	V <sub>DS</sub> =800V, I <sub>D</sub> =35A, Gate		37		
Rise time	t <sub>r</sub>	Driver =-5V to +12V,		12		
Turn-off delay time	t <sub>d(off)</sub>	Turn-on $R_{G,EXT}$ =8.5 $\Omega$ , Turn-off $R_{G,EXT}$ =22 $\Omega$ Inductive Load, FWD: same device with		47		ns
Fall time	t <sub>f</sub>			7		
Turn-on energy	E <sub>ON</sub>			670		
Turn-off energy	E <sub>OFF</sub>	$V_{GS} = -5V, R_{G} = 22\Omega,$		129		μJ
Total switching energy	E <sub>TOTAL</sub>	T <sub>J</sub> =150°C		799		





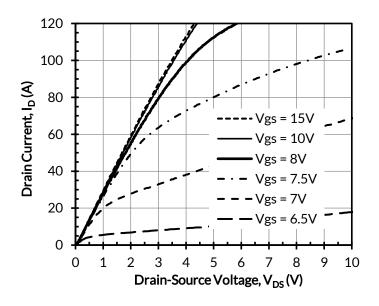








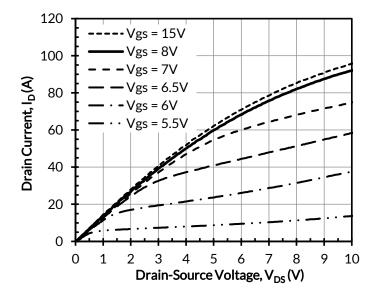
### **Typical Performance Diagrams**



120 100 Drain Current, I<sub>D</sub> (A) 80 Vgs = 15V Vgs = 10V 60 Vgs = 8V 40 Vgs = 7VVgs = 6.5V20 Vgs = 6V 0 1 2 10 Drain-Source Voltage, V<sub>DS</sub> (V)

Figure 1. Typical output characteristics at  $T_J = -55$ °C, tp < 250 $\mu$ s

Figure 2. Typical output characteristics at  $T_J = 25$ °C,  $tp < 250\mu s$ 



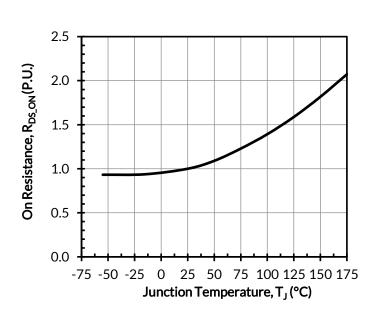


Figure 3. Typical output characteristics at  $T_J$  = 175°C, tp < 250 $\mu$ s

Figure 4. Normalized on-resistance vs. temperature at  $V_{GS}$  = 12V and  $I_D$  = 35A



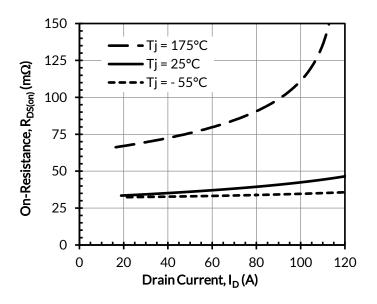








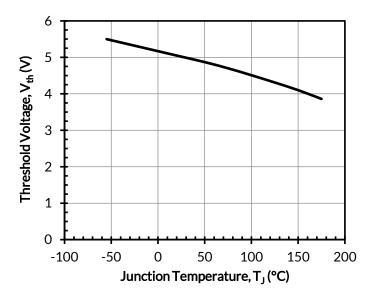




Tj = -55°C Tj = 25°C Drain Current, I<sub>D</sub> (A) Tj = 175°C Gate-Source Voltage, V<sub>GS</sub> (V)

Figure 5. Typical drain-source on-resistances at  $V_{\text{GS}}$  = 12V

Figure 6. Typical transfer characteristics at  $V_{DS} = 5V$ 



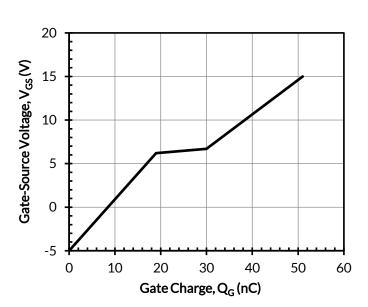


Figure 7. Threshold voltage vs. junction temperature at  $V_{DS}$  = 5V and  $I_{D}$  = 10mA

Figure 8. Typical gate charge at  $V_{DS}$  = 800V and  $I_{D}$  = 35A





0









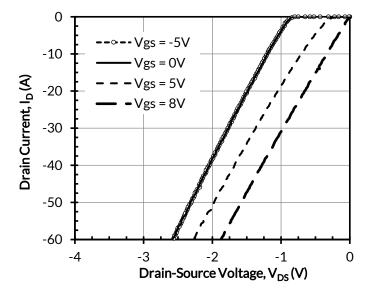
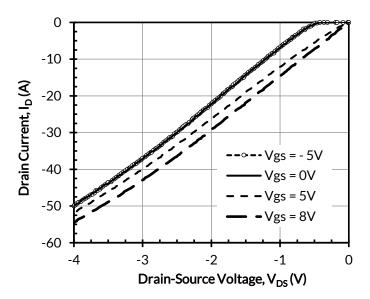


Figure 9. 3rd quadrant characteristics at  $T_J = -55$ °C

Figure 10. 3rd quadrant characteristics at  $T_J = 25$ °C



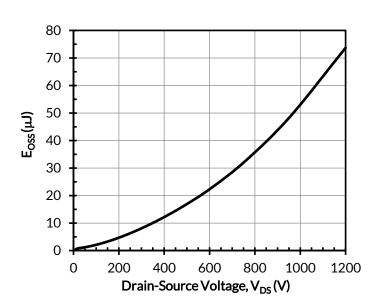


Figure 11. 3rd quadrant characteristics at  $T_J = 175$ °C

Figure 12. Typical stored energy in  $C_{OSS}$  at  $V_{GS} = 0V$ 



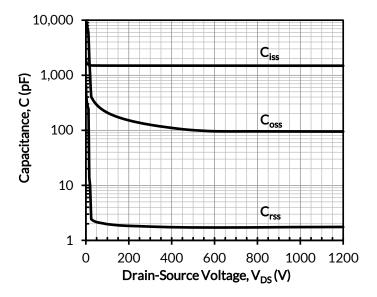








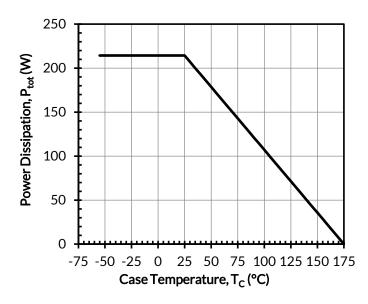




30 40 40 35 30 25 20 15 10 -75 -50 -25 0 25 50 75 100 125 150 175 Case Temperature, T<sub>C</sub> (°C)

Figure 13. Typical capacitances at f = 100kHz and  $V_{GS}$  = 0V

Figure 14. DC drain current derating



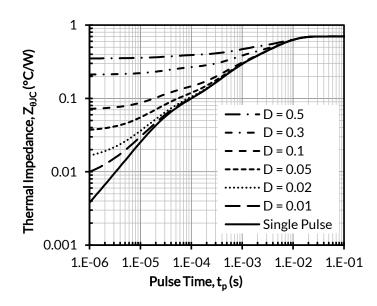


Figure 15. Total power dissipation

Figure 16. Maximum transient thermal impedance















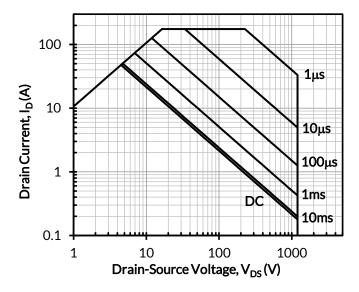


Figure 17. Safe operation area at  $T_C$  = 25°C, D = 0, Parameter  $t_p$ 

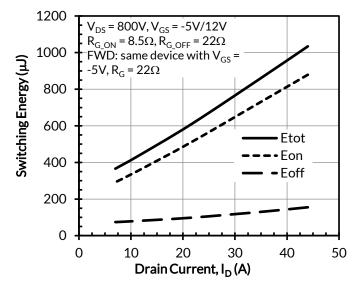


Figure 18. Clamped inductive switching energy vs. drain current at  $T_J = 25$ °C

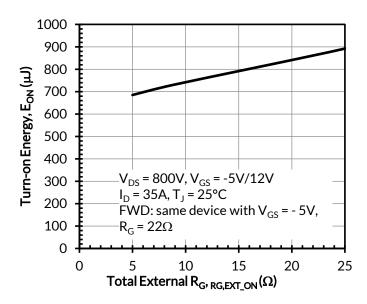


Figure 19. Clamped inductive switching turn-on energy vs.  $R_{G,\text{EXT\_ON}}$ 

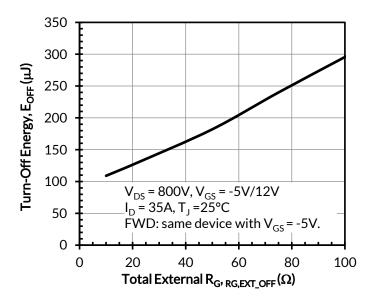


Figure 20. Clamped inductive switching turn-off energy vs.  $R_{G,EXT\ OFF}$ 



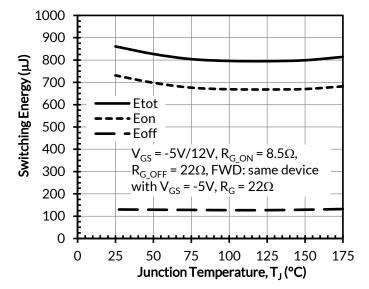












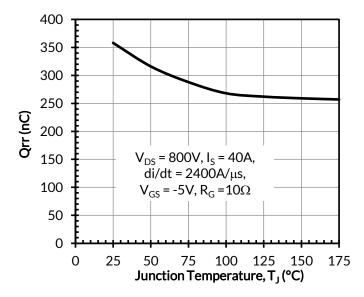


Figure 21. Clamped inductive switching energy vs. junction temperature at  $V_{DS}$  = 800V and  $I_D$  = 35A

Figure 22. Reverse recovery charge Qrr vs. junction temperature

### **Applications Information**

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ( $R_{DS(on)}$ ), output capacitance ( $C_{oss}$ ), gate charge ( $C_{oss}$ ), and reverse recovery charge ( $C_{oss}$ ) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small  $R_{(G)}$ , or gate resistor, provides better EMI suppression with higher efficiency compared to using a high  $R_{(G)}$  value. There is no extra gate delay time when using the snubber circuitry, and a small  $R_{(G)}$  will better control both the turn-off  $V_{(DS)}$  peak spike and ringing duration, while a high  $R_{(G)}$  will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high  $R_{(G)}$ , while greatly reducing  $E_{(OFF)}$  from mid-to-full load range with only a small increase in  $E_{(ON)}$ . Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com













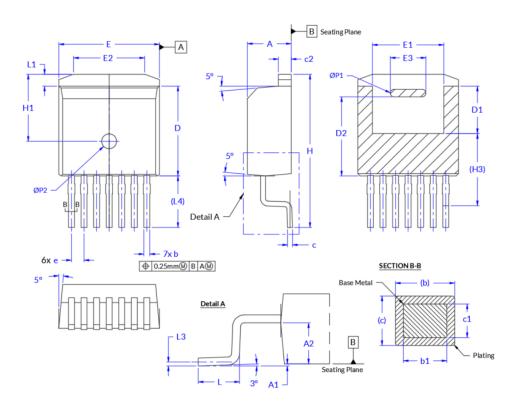
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TO263-7L (D2PAK-7L) PACKAGE OUTLINE, MARKING, TAPE AND REEL SPECIFICATION	PART	Page <b>1</b> of <b>4</b>
DS TO 263 71		Rev D

#### **PACKAGE OUTLINE**



	7L-D2PAK				
SYM	М	М	IN	CH	
31141	Min	Max	Min	Max	
Α	4.30	4.56	.169	.180	
A1	0.00	0.25	.000	.010	
A2	2.45	2.75	.096	.108	
b	0.50	0.70	.020	.028	
b1	0.50	-	.020	-	
С	0.40	0.60	.016	.024	
c1	0.40		.016		
c2	1.20	1.40	.047	.055	
D	8.93	9.23	.352	.363	
D1	4.65	4.95	.183	.195	
D2	7.90	8.10	.311	.319	
e	1.27	BSC	.050 BSC		
E	10.08	10.28	.397	.405	
E1	6.82	7.62	.269	.300	
E2	6.50	8.60	.256	.339	
E3	3.50	3.70	.138	.146	
Н	15.00	16.00	.591	.630	
H1	6.68	6.88	.263	.271	
H3	7.31	REF.	.287	REF	
L	1.90	2.50	.075	.098	
L1	0.98	1.42	.039	.056	
L3	0.25	BSC	.0098	BSC	
L4	5.22	REF	.205	REF	
ØP1	0.65	0.85	.026	.033	
ØP2	1.40	1.60	.055	.063	

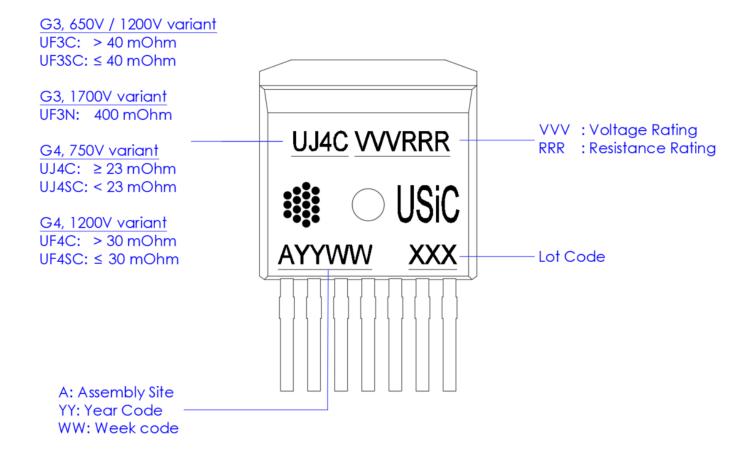
#### Notes:

- 1. GENERAL TOLERANCE: ±0.1 unless otherwise specified
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. PACKAGE BODY SIDES EXCLUDE MOLD FLASH AND GATE BURRS.
- 4. DIMENSION LIS MEASURED IN GAUGE LINE.
- 5. CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
- 6. DIMENSION c1 AND b1 APPLIES TO BASE METAL ONLY



TO263-7L (D2PAK-7L) PACKAGE OUTLINE, PART MARKING, TAPE AND REEL SPECIFICATION	Page <b>2</b> of <b>4</b>
DS_TO_263_7L	Rev D

#### **PART MARKING**



Template: FOR-000530 Rev G



TO263-7L	(D2PAK-7L)	PACKAGE	OUTLINE,	PART
MARKING,	TAPE AND RE	<b>EL SPECIFIC</b>	ATION	

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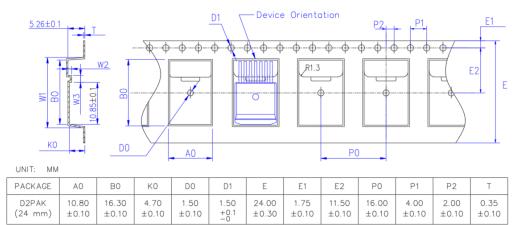
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Rev D

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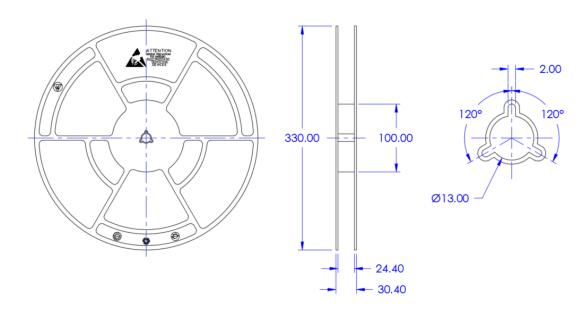
#### **PACKING TYPE**

### Carrier Tape



Ext	erior	size	
	W1	16.9±0.1	
Spec	W2	1.3±0.1	
'	W3	1.0±0.1	
	W1	17.2±0.1	(1)
Spec 2	W2	1.8±0.1	<b>(b)</b>
	W3	0.85±0.1	0

#### Reel



All dimensions in millimeters Anti-Static Tape and Rell (T&R) Quantity per Reel: 800 units



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#### **REVISION HISTORY**

Revision	Create Date (mm/dd/yyyy)	Description of Change	Initiator of Change
С	11/06/2023	Updated to Qorvo template Updated Package outline drawing based latest drawing revision	Glenn Galang
D	05/21/2024	Added illustration of device orientation on carrier tape (page 3)	Glenn Galang

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