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SiC JFET Division

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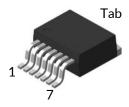
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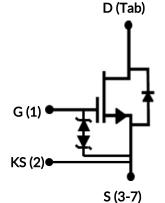


Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, D2PAK-7L, 1200 V, 23 mohm

DATASHEET

UF4SC120023B7S





Part Number	Package	Marking
UF4SC120023B7S	D ² PAK-7L	UF4SC120023B7S



Rev. B, January 2025

Description

The UF4SC120023B7S is a 1200V, $23m\Omega$ G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal redesign when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving D²PAK-7L package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance $R_{DS(on)}$: 23m Ω (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q_{rr} = 243 nC
- Low body diode V_{FSD}: 1.2V
- Low gate charge: Q_G = 37.8nC
- Threshold voltage V_{G(th)}: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected: HBM class 2 and CDM class C3
- D²PAK-7L package for faster switching, clean gate waveforms
- AECQ Qualified

Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Induction heating





Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V _{DS}		1200	V
Gate-source voltage	V _{GS}	DC	-20 to +20	V
Continuous drain current ¹	1	T _c = 25°C	72	А
	ID	T _C = 100°C	51	А
Pulsed drain current ²	I _{DM}	T _c = 25°C	204	А
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =4.1A	126	mJ
SiC FET dv/dt ruggedness	dv/dt	V _{DS} ≤ 800V	150	V/ns
Power dissipation	P _{tot}	T _C = 25°C	385	W
Maximum junction temperature	T _{J,max}		175	°C
Operating and storage temperature	T _J , T _{STG}		-55 to 175	°C
Reflow soldering Temperature	T _{solder}	reflow MSL 1	245	°C

1. Limited by $T_{J,max}$

2. Pulse width $t_{\rm p}$ limited by $T_{\rm J,max}$

3. Starting T_J = 25°C

Thermal Characteristics

Parameter	Symbol Test Conditions			Units		
Parameter	Symbol		Min	Тур	Max	Units
Thermal resistance, junction-to-case	$R_{\theta JC}$			0.3	0.39	°C/W





Electrical Characteristics (T_J = +25°C unless otherwise specified)

Typical Performance - Static

Parameter	Symbol	Test Conditions		Linita			
Parameter	Symbol	Test Conditions	Min	Тур	Max	- Units	
Drain-source breakdown voltage	BV _{DS}	V _{GS} =0V, I _D =1mA	1200			V	
		V _{DS} =1200V,		0	(0		
Total drain lookago current		V _{GS} =0V, T _J =25°C		2	60	- μΑ	
Total drain leakage current	I _{DSS}	V _{DS} =1200V,		20			
		V _{GS} =0V, T _J =175°C		20			
	1	V _{DS} =0V, T _J =25°C,		/	100		
Total gate leakage current	I _{GSS}	V _{GS} =-20V / +20V		6	±20	μΑ	
		V _{GS} =12V, I _D =40A,		22	20		
		TJ=25°C		23	30		
		V _{GS} =12V, I _D =40A,		40		mΩ	
Drain-source on-resistance	R _{DS(on)}	T _J =125°C		42			
		V _{GS} =12V, I _D =40A,		(0			
		T _J =175°C		62			
Gate threshold voltage	V _{G(th)}	V _{DS} =5V, I _D =10mA	4	4.8	6	V	
Gate resistance	R _G	f=1MHz, open drain		4.5		Ω	

Typical Performance - Reverse Diode

Parameter	Symbol Test Conditions			Units			
Parameter	Symbol	Test Conditions	Min	Тур	Max	- Offics	
Diode continuous forward current ¹	ا _s	Т _с = 25°С			72	А	
Diode pulse current ²	I _{S,pulse}	T _C = 25°C			204	А	
Forward voltage	V_{FSD}	V _{GS} =0V, I _S =20A, T _J =25°C		1.2	1.4	v	
Forward voltage	♥ FSD	V _{GS} =0V, I _S =20A, T _J =175°C		1.65			
Reverse recovery charge	Q _{rr}	V _R =800V, I _S =40A, V _{GS} =0V, R _G =50Ω		243		nC	
Reverse recovery time	t _{rr}	di/dt=2000A/μs, Τ _J =25°C		26.8		ns	
Reverse recovery charge	Q _{rr}	V _R =800V, I _S =40A, V _{GS} =0V, R _G =50Ω		264		nC	
Reverse recovery time	t _{rr}	di/dt=2000A/µs, T_=150°C		28.8		ns	





Typical Performance - Dynamic

Demonstern	Course have a	Test Canditians	Value			1.1
Parameter	Symbol	Test Conditions	Min	Тур	Max	Units
Input capacitance	C _{iss}	V _{DS} =800V, V _{GS} =0V		1430		
Output capacitance	C _{oss}	f=100kHz		85		pF
Reverse transfer capacitance	C _{rss}			2		
Effective output capacitance, energy	C	V _{DS} =0V to 800V,		108		pF
related	C _{oss(er)}	V _{GS} =0V		100		рг
Effective output capacitance, time	C	V_{DS} =0V to 800V,		200		pF
related	C _{oss(tr)}	V _{GS} =0V		200		рі
C _{OSS} stored energy	E _{oss}	V_{DS} =800V, V_{GS} =0V		35		μJ
Total gate charge	Q_{G}	- V _{DS} =800V, I _D =40A, -		37.8		-
Gate-drain charge	Q_{GD}	$V_{GS} = 0V \text{ to } 15V$		8		nC
Gate-source charge	Q _{GS}			11.8		
Turn-on delay time	t _{d(on)}	Note 4 and 5,		23		
Rise time	t _r	V _{DS} =800V, I _D =40A, Gate		25		ns
Turn-off delay time	t _{d(off)}	Driver =0V to +15V, R_{G_ON} =10 Ω , R_{G_OFF} =20 Ω ,		64		
Fall time	t _f			10		
Turn-on energy including R_s energy	E _{ON}	inductive Load,		719		
Turn-off energy including R_s energy	E _{OFF}	FWD: same device with $V_{GS} = 0V$ and $R_{G} = 20\Omega$,		95		
Total switching energy	E _{TOTAL}	Snubber: $R_s=10\Omega$,		814		mJ
Snubber R_s energy during turn-on	E _{RS_ON}	C _s =100pF		8		
Snubber R_s energy during turn-off	E_{RS_OFF}	T _J =25°C		15		
Turn-on delay time	t _{d(on)}	Note 4 and 5,		21		
Rise time	t _r	V _{DS} =800V, I _D =40A, Gate		27		nc
Turn-off delay time	t _{d(off)}	Driver =0V to +15V,		63		ns
Fall time	t _f	$R_{G_ON}=10\Omega$, $R_{G_OFF}=20\Omega$,		10		
Turn-on energy including R_s energy	E _{ON}	 inductive Load, FWD: same device with 		781		
Turn-off energy including R_s energy	E _{OFF}	$V_{GS} = 0V$ and $R_G = 20\Omega$,		111		
Total switching energy	E _{TOTAL}	Snubber: $R_s = 10\Omega$,		892		μJ
Snubber R_s energy during turn-on	E _{RS_ON}	C _s =100pF		11		
Snubber R_s energy during turn-off	E_{RS_OFF}	T _J =150°C		15		

4. Measured with the switching test circuit in Figure 26.

5. In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.

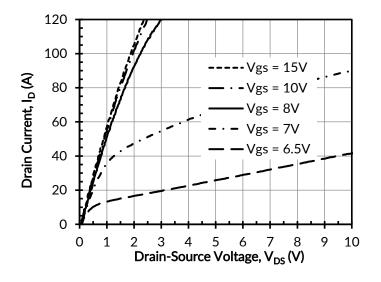


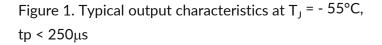
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Typical Performance Diagrams





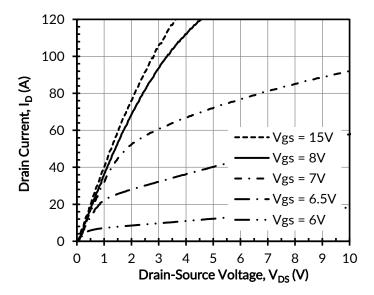


Figure 2. Typical output characteristics at $T_J = 25^{\circ}$ C, tp < 250 μ s

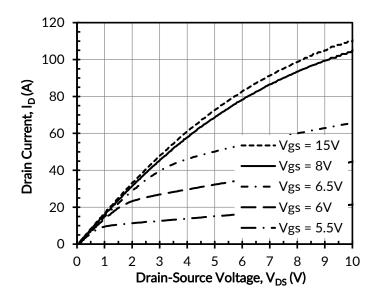


Figure 3. Typical output characteristics at $T_J = 175^{\circ}C$, tp < 250 μ s

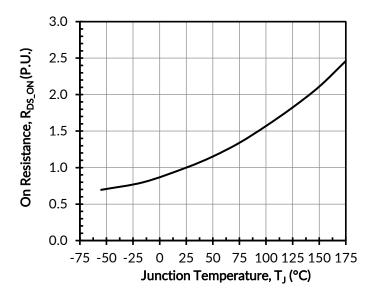
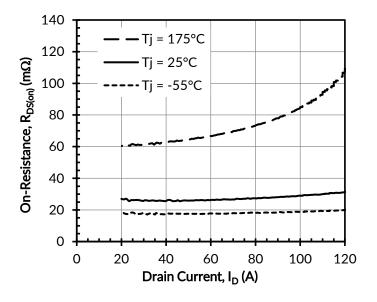
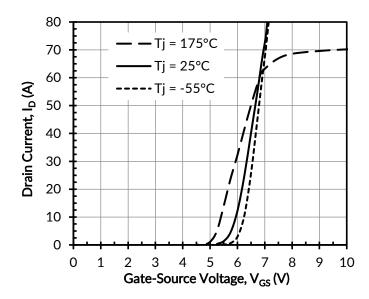


Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_D = 40A





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Figure 5. Typical drain-source on-resistances at V_{GS} = 12V

Figure 6. Typical transfer characteristics at V_{DS} = 5V

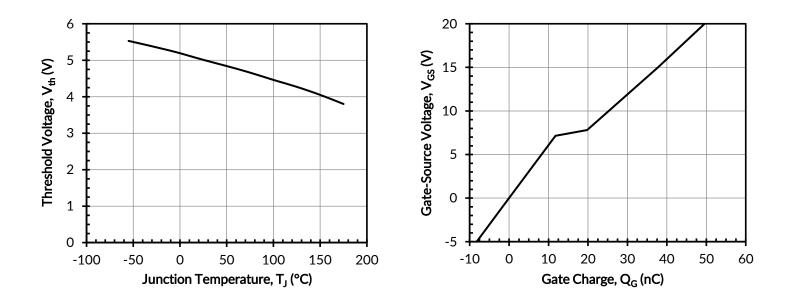


Figure 7. Threshold voltage vs. junction temperature at Figure 8. Typical gate charge at at V_{DS} = 800V I_D = 40A V_{DS} = 5V and I_{D} = 10mA

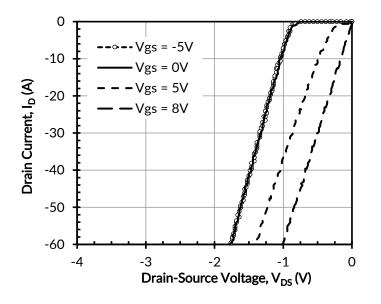
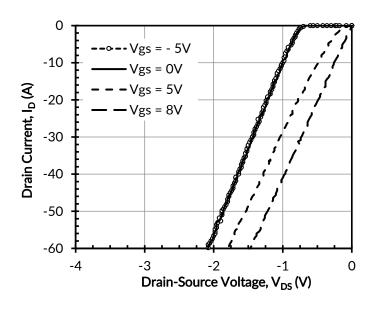


Figure 9. 3rd quadrant characteristics at $T_J = -55^{\circ}C$



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Figure 10. 3rd quadrant characteristics at $T_J = 25^{\circ}C$

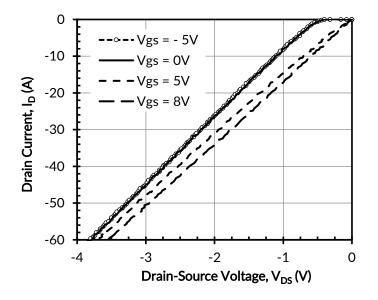


Figure 11. 3rd quadrant characteristics at $T_J = 175^{\circ}C$

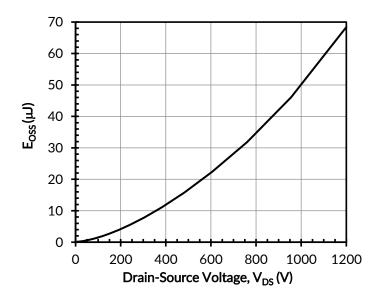
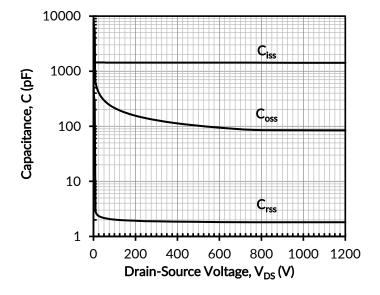
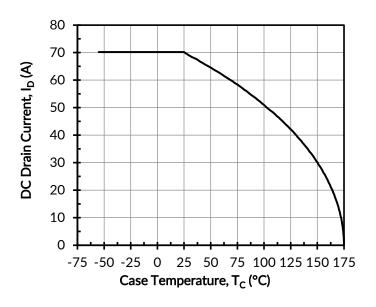


Figure 12. Typical stored energy in C_{OSS} at V_{GS} = 0V





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Figure 13. Typical capacitances at f = 100kHz and V_{GS} = 0V

Figure 14. DC drain current derating

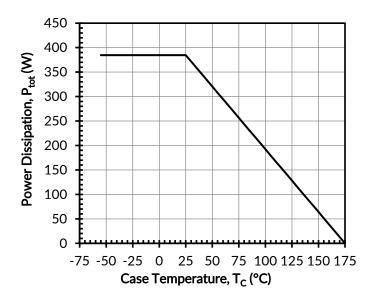


Figure 15. Total power dissipation

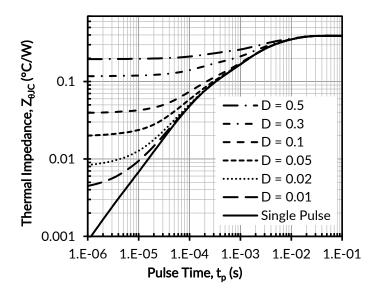


Figure 16. Maximum transient thermal impedance

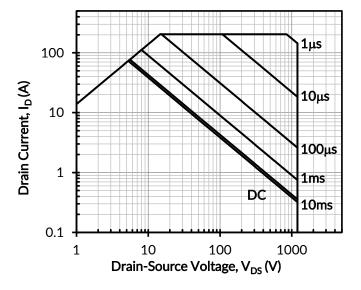
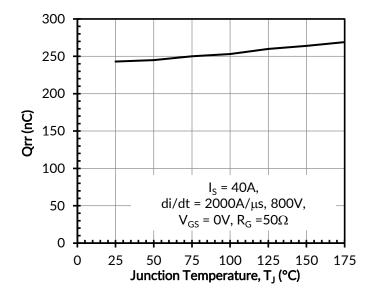


Figure 17. Safe operation area at T_C = 25°C, D = 0, Parameter t_p



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Figure 18. Reverse recovery charge Qrr vs. junction temperature at V_{DS} = 800V

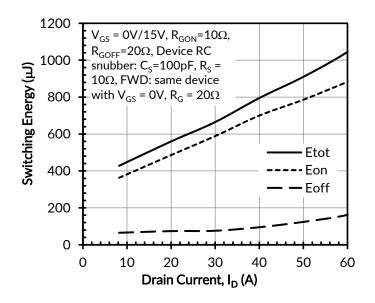


Figure 19. Clamped inductive switching energy vs. drain current at V_{DS} = 800V and T_J = 25°C

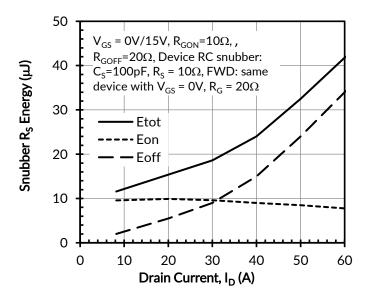
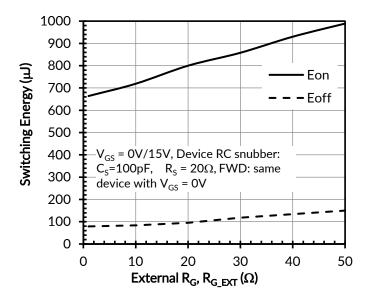
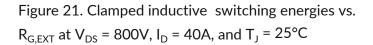


Figure 20. RC snubber energy loss vs. drain current at V_{DS} = 800V and T_J = 25°C





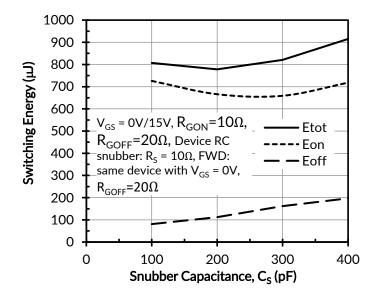
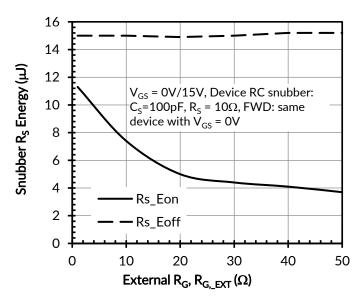


Figure 23. Clamped inductive switching energies vs. snubber capacitance C_S at V_{DS} = 800V, I_D = 40A, and T_J = 25°C



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Figure 22. RC snubber energy loss vs. $R_{G,EXT}$ at V_{DS} = 800V, I_D = 40A, and T_J = 25°C

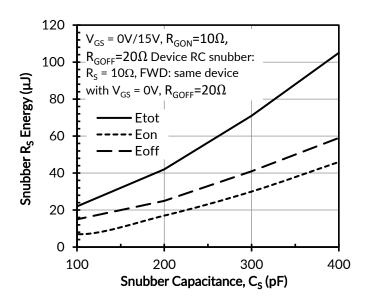
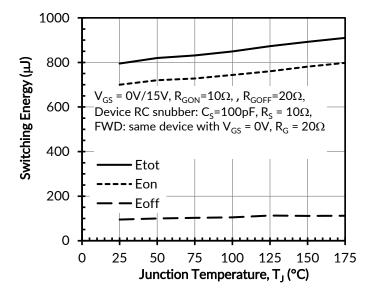
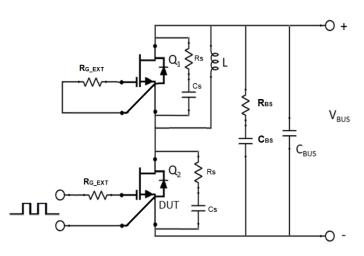


Figure 24. RC snubber energy losses vs. snubber capacitance C_s at V_{DS} = 800V, I_D =40A, and T_J = 25°C





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Figure 25. Clamped inductive switching energy vs. junction temperature at V_{DS} =800V and I_D =40A

Figure 26.Schematic of the half-bridge mode switching test circuit with device RC snubbers (Rs =10 Ω , Cs = 100pF) and a bus RC snubber (R_{BS} = 2.5 Ω , C_{BS}=100nF).

Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance (R_{DS(on)}), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode. Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com



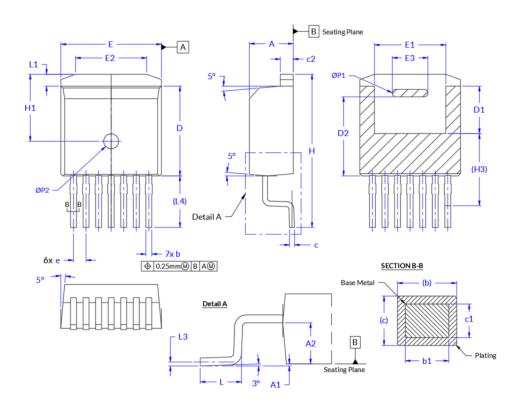


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PACKAGE OUTLINE



	7L-D2PAK						
SYM	M	М	IN	СН			
SIM	Min	Max	Min	Max			
A	4.30	4.56	.169	.180			
A1	0.00	0.25	.000	.010			
A2	2.45	2.75	.096	.108			
b	0.50	0.70	.020	.028			
b1	0.50		.020	-			
с	0.40	0.60	.016	.024			
c1	0.40		.016				
c2	1.20	1.40	.047	.055			
D	8.93	9.23	.352	.363			
D1	4.65	4.95	.183	.195			
D2	7.90	8.10	.311	.319			
e	1.27	BSC	.050	BSC			
E	10.08	10.28	.397	.405			
E1	6.82	7.62	.269	.300			
E2	6.50	8.60	.256	.339			
E3	3.50	3.70	.138	.146			
н	15.00	16.00	.591	.630			
H1	6.68	6.88	.263	.271			
H3	7.3	REF.	.287	REF			
L	1.90	2.50	.075	.098			
L1	0.98	1.42	.039	.056			
L3	0.25	BSC	.0098	BSC			
L4	5.22	REF	.205	REF			
ØP1	0.65	0.85	.026	.033			
ØP2	1.40	1.60	.055	.063			

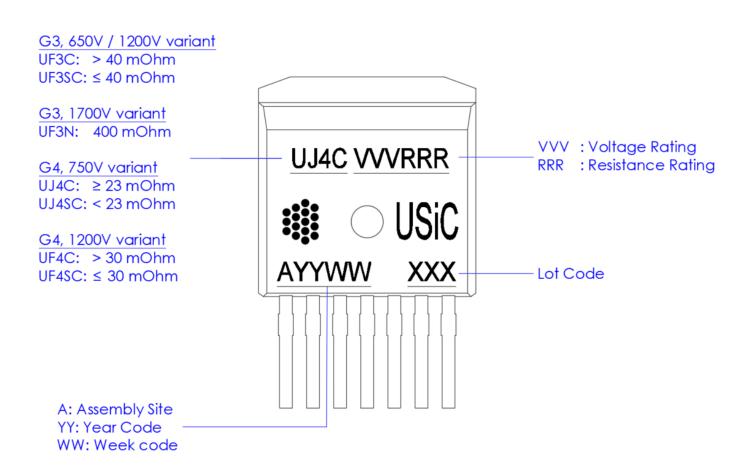
Notes:

- 1. GENERAL TOLERANCE: ±0.1 unless otherwise specified
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. PACKAGE BODY SIDES EXCLUDE MOLD FLASH AND GATE BURRS.
- 4. DIMENSION L IS MEASURED IN GAUGE LINE.
- 5. CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
- 6. DIMENSION c1 AND b1 APPLIES TO BASE METAL ONLY



TO263-7L (D2PAK-7L) PACKAGE OUTLINE, PA MARKING, TAPE AND REEL SPECIFICATION	ART	Page 2 of 4
DS_TO_263_7L		Rev D

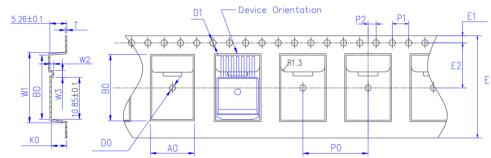
PART MARKING





PACKING TYPE

Carrier Tape

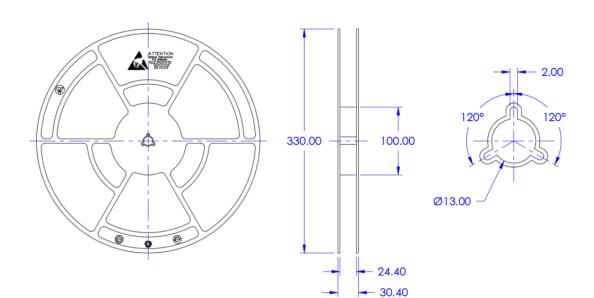


UNIT: MM

PACKAGE	AO	BO	KO	DO	D1	E	E1	E2	P0	P1	P2	Т
D2PAK (24 mm)	10.80 ±0.10	16.30 ±0.10	4.70 ±0.10	1.50 ±0.10	1.50 +0.1 -0	24.00 ±0.30	1.75 ±0.10	11.50 ±0.10	16.00 ±0.10	4.00 ±0.10	2.00 ±0.10	0.35 ±0.10

Exte	erior	size	
-	W1	16.9±0.1	
Spec 1	W2	1.3±0.1	
	W3	1.0±0.1	
-	W1	17.2±0.1	0
Spec 2	W2	1.8±0.1	Б
2	W3	0.85±0.1	\bigcirc

<u>Reel</u>



All dimensions in millimeters Anti-Static Tape and Rell (T&R) Quantity per Reel: 800 units



TO263-7L (D2PAK-7L) PACKAGE OUTLINE, MARKING, TAPE AND REEL SPECIFICATION	PART	Page 4 of 4
DS_TO_263_7L		Rev D

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REVISION HISTORY

Revision	Create Date (mm/dd/yyyy)	Description of Change	Initiator of Change
С	11/06/2023	Updated to Qorvo template Updated Package outline drawing based latest drawing revision	Glenn Galang
D	05/21/2024	Added illustration of device orientation on carrier tape (page 3)	Glenn Galang

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