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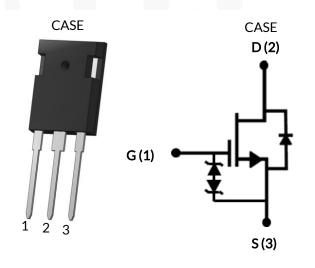
Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, TO-247-3L, 750 V, 58 mohm

More

Rev. B, January 2025

DATASHEET

UJ4C075060K3S



Part Number	Package	Marking
UJ4C075060K3S	TO-247-3L	UJ4C075060K3S



Description

The UJ4C075060K3S is a 750V, $58m\Omega$ G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the TO-247-3L package, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance $R_{DS(on)}$: 58m Ω (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q_{rr} = 52nC
- Low body diode V_{FSD}: 1.31V
- Low gate charge: Q_G = 37.8nC
- Threshold voltage V_{G(th)}: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected, HBM class 2
- AECQ Qualified

Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating





Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V _{DS}		750	V
Gate-source voltage	V _{GS}	DC	-20 to +20	V
Continuous drain current ¹		T _C = 25°C	28	А
Continuous drain current	ID	T _C = 100°C	20.6	А
Pulsed drain current ²	I _{DM}	T _C = 25°C	62	А
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =1.8A	24.3	mJ
Power dissipation	P _{tot}	T _C = 25°C	155	W
Maximum junction temperature	T _{J,max}		175	°C
Operating and storage temperature	T _J , T _{STG}		-55 to 175	°C
Max. lead temperature for soldering, 1/8" from case for 5 seconds	ΤL		250	°C

1. Limited by $T_{J,max}$

2. Pulse width t_p limited by $T_{J,max}$

3. Starting $T_J = 25^{\circ}C$

Thermal Characteristics

Deremeter	Symbol	Test Conditions	Value			Linite
Parameter			Min	Тур	Max	Units
Thermal resistance, junction-to-case	$R_{ ext{ heta}JC}$			0.75	0.97	°C/W









Electrical Characteristics (T_J = +25°C unless otherwise specified)

Typical Performance - Static

Parameter	Symbol	Test Conditions	Value			Unite
			Min	Тур	Max	- Units
Drain-source breakdown voltage	BV _{DS}	V_{GS} =0V, I_{D} =1mA	750			V
Total drain leakage current	I _{DSS}	V _{DS} =750V, V _{GS} =0V, T _J =25°C		0.7	40	μΑ
		V _{DS} =750V, V _{GS} =0V, T _J =175°C		15		
Total gate leakage current	I _{GSS}	V _{DS} =0V, T _J =25°C, V _{GS} =-20V / +20V		4.7	±20	μA
Drain-source on-resistance	R _{DS(on)}	V _{GS} =12V, I _D =20A, T _J =25°C		58	74	
		V _{GS} =12V, I _D =20A, T _J =125°C		106		mΩ
		V _{GS} =12V, I _D =20A, T _J =175°C		147		1
Gate threshold voltage	V _{G(th)}	V_{DS} =5V, I_{D} =10mA	4	4.8	6	V
Gate resistance	R _G	f=1MHz, open drain		4.5		Ω

Typical Performance - Reverse Diode

Parameter	Symbol	Test Conditions	Value			- Units
Parameter			Min	Тур	Max	Units
Diode continuous forward current ¹	ls	T _C =25°C			28	А
Diode pulse current ²	I _{S,pulse}	T _C =25°C			62	А
Forward voltage	V _{FSD}	V _{GS} =0V, I _F =10A, T _J =25°C		1.31	1.75	- V
		V _{GS} =0V, I _F =10A, T _J =175°C		1.8		
Reverse recovery charge	Q _{rr}	V_{R} =400V, I_{F} =20A, V_{GS} =0V, $R_{G_{EXT}}$ =20 Ω		52		nC
Reverse recovery time	t _{rr}	di/dt=1060A/μs, Tյ=25°C		16		ns
Reverse recovery charge	Q _{rr}	V_R =400V, I_F =20A, V_{GS} =0V, R_{G_EXT} =20 Ω		58		nC
Reverse recovery time	t _{rr}	di/dt=1060A/µs, T_=150°C		19		ns







Typical Performance - Dynamic

Parameter	Symbol	Test Conditions	Value			Units
Parameter			Min	Тур	Max	Units
Input capacitance	C _{iss}	V _{DS} =100V, V _{GS} =0V		1422		
Output capacitance	C _{oss}	$v_{DS} = 100 \text{ V}, v_{GS} = 0 \text{ V}$ f=100kHz		68		pF
Reverse transfer capacitance	C _{rss}	1-100K112		2.7		
Effective output capacitance, energy related	C _{oss(er)}	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		50		pF
Effective output capacitance, time related	C _{oss(tr)}	V_{DS} =0V to 400V, V_{GS} =0V		94		pF
C _{OSS} stored energy	E _{oss}	V _{DS} =400V, V _{GS} =0V		4		μJ
Total gate charge	Q _G	- V _{DS} =400V, I _D =20A, -		37.8		
Gate-drain charge	Q_{GD}	$V_{DS} = 400 V, T_D = 20 A,$ $V_{GS} = 0V \text{ to } 15 V$		8		nC
Gate-source charge	Q _{GS}	VGS 0V 10 13 V		11.8		
Turn-on delay time	t _{d(on)}	Note 4,		13		
Rise time	t _r	V_{DS} =400V, I_{D} =20A, Gate		29		nc
Turn-off delay time	$t_{d(off)}$	Driver =0V to +15V, Turn-on $R_{G,EXT}$ =1 Ω ,		78		– ns
Fall time	t _f	Turn-off $R_{G,EXT}$ =20 Ω		13		
Turn-on energy	E _{ON}	Inductive Load, FWD: same device with		168		
Turn-off energy	E _{OFF}	$V_{GS} = 0V, R_G = 20\Omega,$		58		μJ
Total switching energy	E _{TOTAL}	T _J =25°C		226		
Turn-on delay time	t _{d(on)}	Note 4,		13		
Rise time	t _r	V _{DS} =400V, I _D =20A, Gate		31		ns
Turn-off delay time	t _{d(off)}	Driver =0V to +15V, Turn-on $R_{G,EXT}$ =1 Ω ,		84		
Fall time	t _f	Turn-off $R_{G,EXT}$ =20 Ω		14		
Turn-on energy	E _{ON}	Inductive Load, FWD: same device with		189		
Turn-off energy	E _{OFF}	$V_{GS} = 0V, R_G = 20\Omega,$		70		μJ
Total switching energy	E _{TOTAL}	Т _Ј =150°С		259		1

4. Measured with the half-bridge mode switching test circuit in Figure 28.







Typical Performance - Dynamic (continued)

Doromotor	Cymah al	Test Conditions	Value			
Parameter	Symbol	lest Conditions	Min	Тур	Max	- Units
Turn-on delay time	t _{d(on)}	Note 5, V_{DS} =400V, I_D =20A, Gate Driver =0V to +15V, $R_{G,EXT}$ =1 Ω , inductive Load,		13		
Rise time	t _r			31		
Turn-off delay time	t _{d(off)}			31		- ns
Fall time	t _f			9		
Turn-on energy including R _s energy	E _{ON}	FWD: same device with V_{GS}		186		
Turn-off energy including R _s energy	E _{OFF}	= 0V and $R_G = 1\Omega$, RC snubber: $R_{S1}=10\Omega$ and		18		-
Total switching energy	E _{TOTAL}	$C_{s1} = 95 pF,$		204		μJ
Snubber R _s energy during turn-on	E _{RS_ON}	T_=25°C		0.5		_
Snubber R _s energy during turn-off	E _{RS_OFF}			0.9		-
Turn-on delay time	t _{d(on)}	Note 5,		13		
Rise time	t _r			35		- ns
Turn-off delay time	t _{d(off)}	V_{DS} =400V, I_D =20A, Gate		34		
Fall time	t _f	$- \frac{\text{Driver =0V to +15V,}}{\text{R}_{G,EXT}=1\Omega, \text{ inductive Load,}}$		10		
Turn-on energy including R _s energy	E _{ON}	FWD: same device with V_{GS}		210		μ
Turn-off energy including R _s energy	E _{OFF}	= 0V and $R_G = 1\Omega$, RC		24		
Total switching energy	E _{TOTAL}	snubber: R_{s1} =10 Ω and C_{s1} =95pF,		234		
Snubber R _s energy during turn-on	E _{RS_ON}	T_=150°C		0.5		
Snubber R _s energy during turn-off	E _{RS_OFF}			0.9		
Turn-on delay time	t _{d(on)}	Note 6,		13		– ns
Rise time	t _r	V _{DS} =400V, I _D =20A, Gate		26		
Turn-off delay time	t _{d(off)}	Driver = 0V to +15V,		78		
Fall time	t _f	Turn-on $R_{G,EXT} = 1\Omega$,		12		
Turn-on energy	E _{ON}	$- Turn-off R_{G,EXT}=20\Omega$ $- Inductive Load,$		142		
Turn-off energy	E _{OFF}	FWD: UJ3D06510TS		56		μJ
Total switching energy	E _{TOTAL}	T_=25°C		198		-
Turn-on delay time	t _{d(on)}	Note 6,		13		
Rise time	t _r	V _{DS} =400V, I _D =20A, Gate		30		1
Turn-off delay time	t _{d(off)}	Driver =0V to +15V,		83		ns
Fall time	t _f	Turn-on $R_{G,EXT} = 1\Omega$,		15		
Turn-on energy	E _{ON}	$- \text{Turn-off } R_{G,EXT} = 20\Omega - $ Inductive Load,		162		
Turn-off energy	E _{OFF}	FWD:UJ3D06510TS T_=150°C		70		μJ
Total switching energy	E _{TOTAL}			232		

5. Measured with the chopper mode switching test circuit in Figure 30.

6. Measured with the chopper mode switching test circuit in Figure 29.





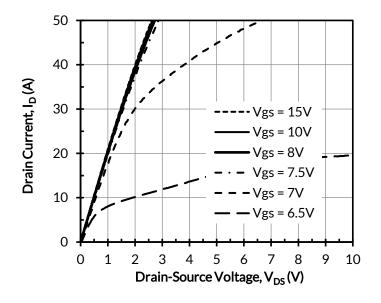
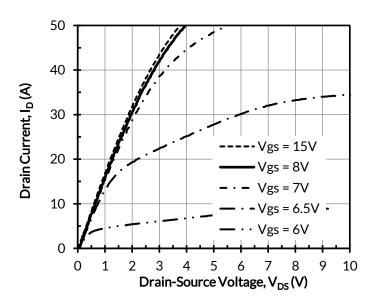


Figure 1. Typical output characteristics at T $_{\rm J}$ = - 55°C, tp < 250 μs



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Figure 2. Typical output characteristics at T $_{\rm J}$ = 25°C, tp < 250 μs

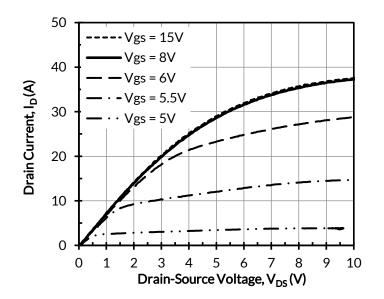


Figure 3. Typical output characteristics at T $_{\rm J}$ = 175°C, tp < 250 μs

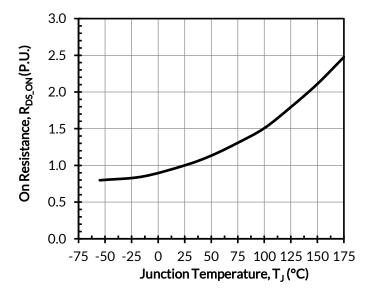


Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_D = 20A





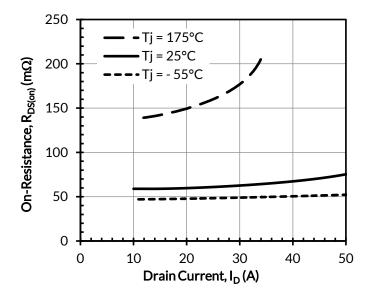


Figure 5. Typical drain-source on-resistances at $V_{\rm GS}$ = 12V

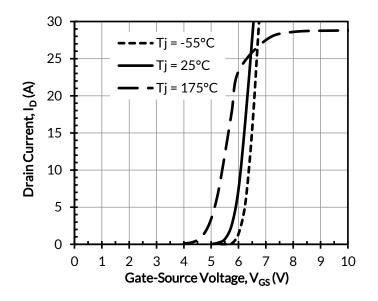


Figure 6. Typical transfer characteristics at V_{DS} = 5V

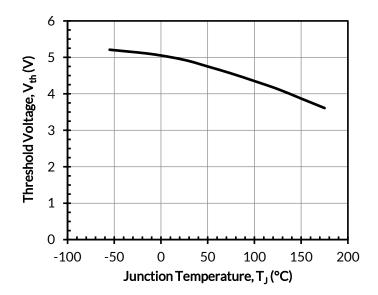


Figure 7. Threshold voltage vs. junction temperature at V_{DS} = 5V and I_{D} = 10mA

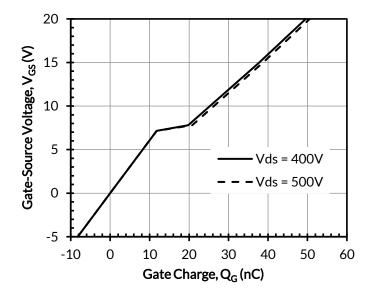


Figure 8. Typical gate charge at I_D = 20A





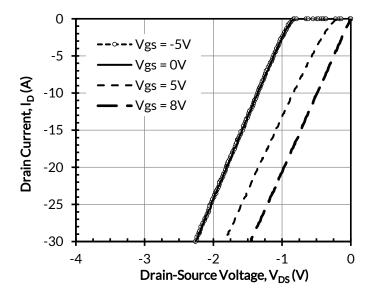


Figure 9. 3rd quadrant characteristics at $T_J = -55^{\circ}C$

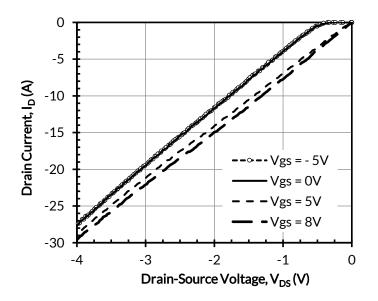


Figure 11. 3rd quadrant characteristics at $T_J = 175^{\circ}C$

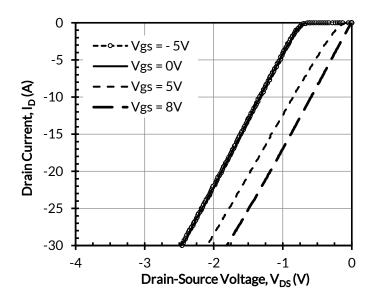


Figure 10. 3rd quadrant characteristics at $T_J = 25^{\circ}C$

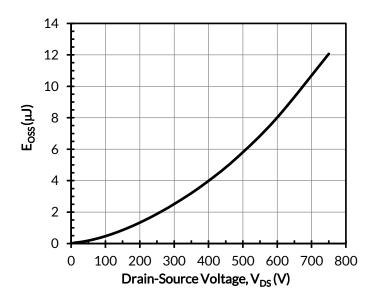


Figure 12. Typical stored energy in C_{OSS} at V_{GS} = 0V





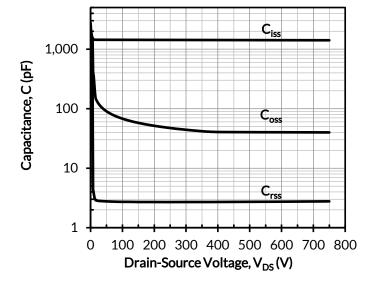


Figure 13. Typical capacitances at f = 100kHz and V_{GS} = 0V

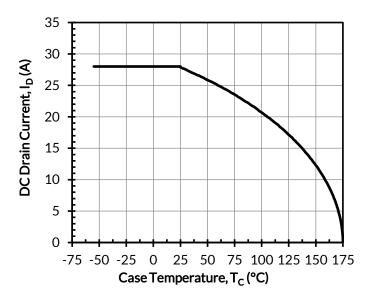


Figure 14. DC drain current derating

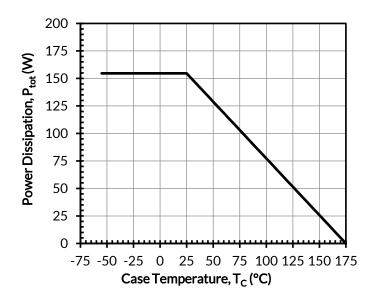


Figure 15. Total power dissipation

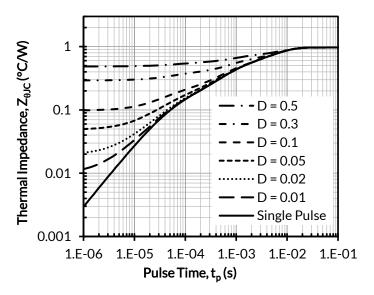


Figure 16. Maximum transient thermal impedance



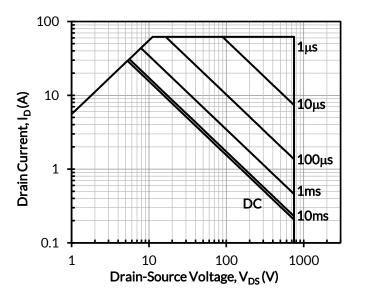
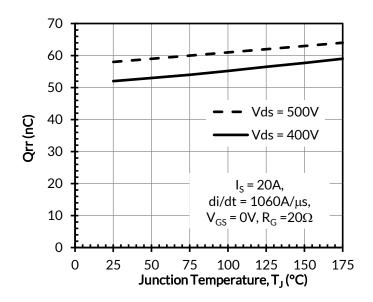


Figure 17. Safe operation area at $T_{\rm C}$ = 25°C, D = 0, Parameter $t_{\rm p}$



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Figure 18. Reverse recovery charge Qrr vs. junction temperature

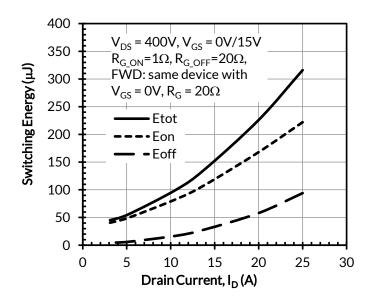


Figure 19. Clamped inductive switching energy vs. drain current at V_{DS} = 400V and T_J = 25°C

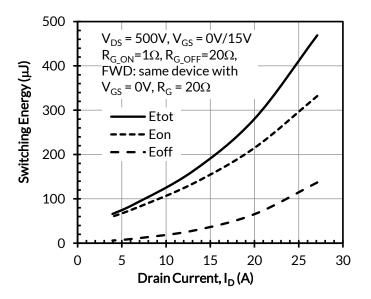


Figure 20. Clamped inductive switching energy vs. drain current at V_{DS} = 500V and T_J = 25°C





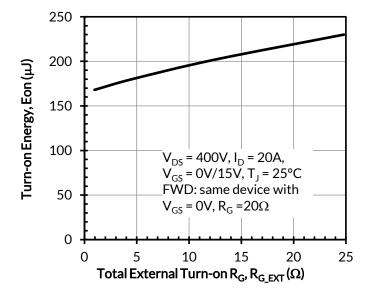


Figure 21. Clamped inductive switching turn-on energy vs. $R_{G,\text{EXT_ON}}$

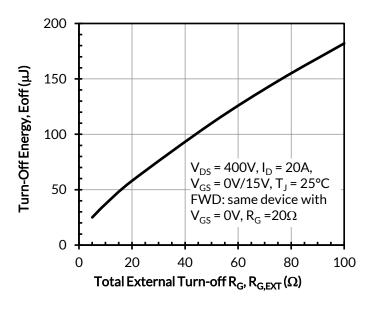


Figure 22. Clamped inductive switching turn-off energy vs. $R_{G,\text{EXT_OFF}}$

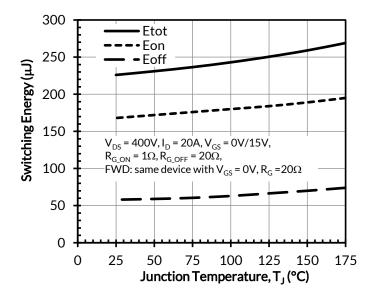


Figure 23. Clamped inductive switching energy vs. junction temperature at V_{DS} =400V and I_D = 20A

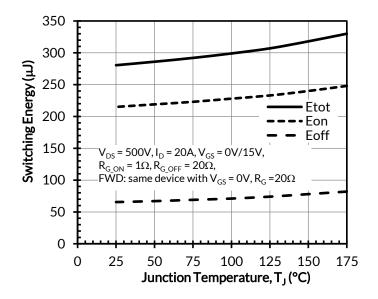


Figure 24. Clamped inductive switching energy vs. junction temperature at V_{DS} =500V and I_D = 20A





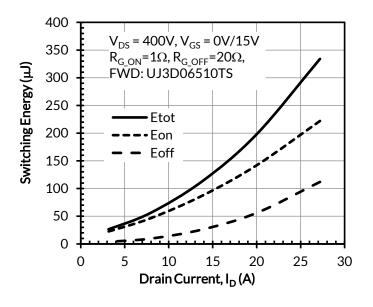


Figure 24. Clamped inductive switching energy vs. drain current at V_{DS} = 400V and T_J = 25°C

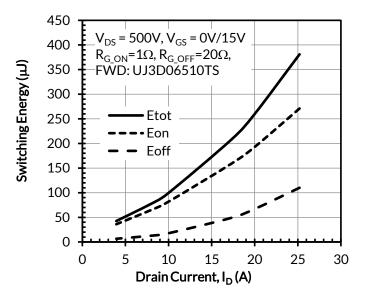


Figure 25. Clamped inductive switching energy vs. drain current at V_{DS} = 500V and $T_{\rm J}$ = 25°C

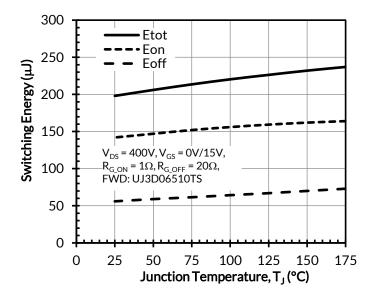


Figure 26. Clamped inductive switching energy vs. junction temperature at V_{DS} =400V and I_D = 20A

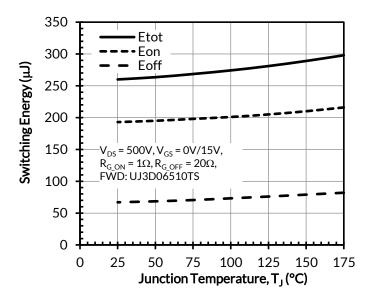


Figure 27. Clamped inductive switching energy vs. junction temperature at V_{DS} =500V and I_D = 20A





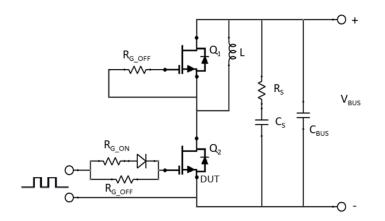


Figure 28. Schematic of the half-bridge mode switching test circuit. Note, a bus RC snubber ($R_s = 2.5\Omega$, $C_s = 100$ nF) is used to reduce the power loop high frequency oscillations.

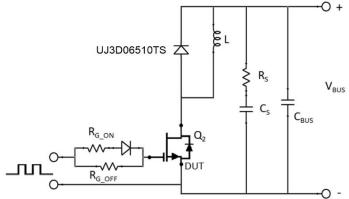


Figure 29. Schematic of the chopper mode switching test circuit. Note, a bus RC snubber ($R_s = 2.5\Omega$, $C_s=100$ nF) is used to reduce the power loop high frequency oscillations.

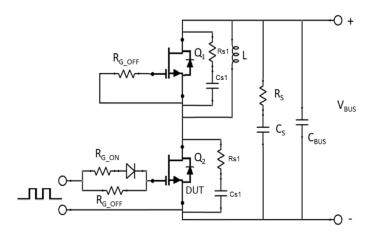


Figure 30. Schematic of the half-bridge mode switching test circuit with device RC snubbers (R_{s1} =10 Ω , C_{s1} = 95pF) and a bus RC snubber (R_{s} = 2.5 Ω , C_{s} =100nF).





Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com

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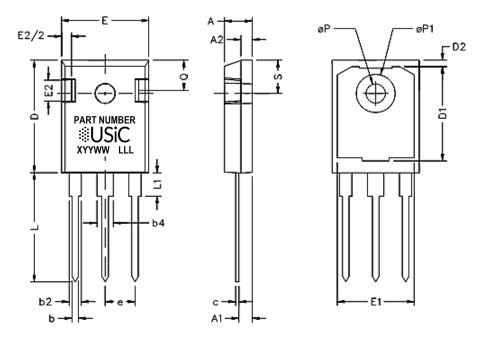
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TO-247-3L PACKAGE OUTLINE, PART MARKING AND TUBE SPECIFICATIONS

PACKAGE OUTLINE



SYM	INC	HES	MILLIN	NETERS
	MIN	MAX	MIN	МАХ
A	0.185	0.209	4.699	5.309
A1	0.087	0.102	2.21	2.61
A2	0.059	0.098	1.499	2.489
b	0.039	0.055	0.991	1.397
b2	0.065	0.094	1.651	2.388
b4	0.102	0.135	2.591	3.429
С	0.015	0.035	0.381	0.889
D	0.819	0.845	20.803	21.463
D1	0.515	-	13.081	-
D2	0.02	0.053	0.508	1.346
E	0.61	0.64	15.494	16.256
е	0.214	4 BSC	5.44	BSC
E1	0.53	-	13.462	-
E2	0.135	0.157	3.429	3.988
L	0.78	0.8	19.812	20.32
L1	-	0.177	-	4.496
ØР	0.14	0.144	3.556	3.658
ØP1	0.278	0.291	7.061	7.391
Q	0.212	0.244	5.385	6.198
S	0.243	3 BSC	6.17	BSC



PART MARKING

PART NUMBER SUSSE XYYWW LLL

PART NUMBER = REFER TO DS_PN DECODER FOR DETAILS

X = ASSEMBLY SITE YY = YEAR WW = WORK WEEK LLL = LOT ID

PACKING TYPE

ANTI-STATIC TUBE

QUANTITY / TUBE : 30 UNITS

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