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### Silicon Carbide (SiC), Cascode JFET -EliteSiC, Power N-Channel, D2PAK-7L, 750V, 18 mohm

Rev. B, January 2025

### Description

The UJ4SC075018B7S is a 750V,  $18m\Omega$  G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the D<sup>2</sup>PAK-7L package, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

### Features

- On-resistance  $R_{DS(on)}$ :  $18m\Omega$  (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q<sub>rr</sub> = 125nC
- Low body diode V<sub>FSD</sub>: 1.14V
- Low gate charge: Q<sub>G</sub> = 37.8nC
- Threshold voltage V<sub>G(th)</sub>: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected, HBM class 2
- D<sup>2</sup>PAK-7L package for faster switching, clean gate waveforms
- AECQ Qualified

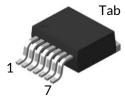
### **Typical applications**

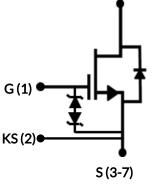
- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating



UJ4SC075018B7S

DATASHEET





D (Tab)

Part Number	Package	Marking
UJ4SC075018B7S	D <sup>2</sup> PAK-7L	UJ4SC075018B7S







### Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V <sub>DS</sub>		750	V
	V	DC	-20 to +20	V
Gate-source voltage	V <sub>GS</sub>	AC (f > 1Hz)	-25 to +25	V
Continuous drain current <sup>1</sup>	1	T <sub>C</sub> = 25°C	72	А
Continuous drain current	Ι <sub>D</sub>	T <sub>C</sub> = 100°C	52	А
Pulsed drain current <sup>2</sup>	I <sub>DM</sub>	T <sub>C</sub> = 25°C	208	А
Single pulsed avalanche energy <sup>3</sup>	E <sub>AS</sub>	L=15mH, I <sub>AS</sub> =3.6A	97.2	mJ
SiC FET dv/dt Ruggedness	$dv/dt_{rug}$	$V_{DS} \le 500V$	200	V/ns
Power dissipation	P <sub>tot</sub>	T <sub>C</sub> = 25°C	259	W
Maximum junction temperature	T <sub>J,max</sub>		175	°C
Operating and storage temperature	Tj, T <sub>STG</sub>		-55 to 175	°C
Reflow soldering temperature	$T_{solder}$	reflow MSL 1	245	°C

1. Limited by  $T_{\mbox{\tiny J,max}}$ 

2. Pulse width  $t_p$  limited by  $T_{J,max}$ 

3. Starting  $T_J = 25^{\circ}C$ 

### **Thermal Characteristics**

Parameter	Symbol Test Conditions			Units		
	Symbol	Test Conditions	Min	Тур	Max	Units
Thermal resistance, junction-to-case	$R_{\theta JC}$			0.45	0.58	°C/W



### Electrical Characteristics (T<sub>J</sub> = +25°C unless otherwise specified)

### **Typical Performance - Static**

Parameter	Cump hal	Test Conditions		Unite			
Parameter	Symbol	Test Conditions	Min	Typ Max		- Units	
Drain-source breakdown voltage	BV <sub>DS</sub>	V <sub>GS</sub> =0V, I <sub>D</sub> =1mA	750			V	
Total duain lookaga guurant		V <sub>DS</sub> =750V, V <sub>GS</sub> =0V, T <sub>J</sub> =25°C		1.3	45		
Total drain leakage current	I <sub>DSS</sub>	<sup>DSS</sup> V <sub>DS</sub> =750V, V <sub>GS</sub> =0V, Т <sub>J</sub> =175°С		20		μA	
Total gate leakage current	I <sub>GSS</sub>	V <sub>DS</sub> =0V, T <sub>J</sub> =25°C, V <sub>GS</sub> =-20V / +20V		4.7	±20	μA	
		V <sub>GS</sub> =12V, I <sub>D</sub> =50A, T <sub>J</sub> =25°C		18	23		
Drain-source on-resistance	R <sub>DS(on)</sub>	V <sub>GS</sub> =12V, I <sub>D</sub> =50A, T <sub>J</sub> =125°C		29		mΩ	
		V <sub>GS</sub> =12V, I <sub>D</sub> =50A, T <sub>J</sub> =175°C		37			
Gate threshold voltage	V <sub>G(th)</sub>	$V_{DS}$ =5V, $I_{D}$ =10mA	4	4.8	6	V	
Gate resistance	R <sub>G</sub>	f=1MHz, open drain		4.5		Ω	

### Typical Performance - Reverse Diode

Parameter	Symbol	Test Conditions		- Units		
	Symbol		Min	Тур	Typ Max	
Diode continuous forward current <sup>1</sup>	ls	T <sub>C</sub> =25°C			72	А
Diode pulse current <sup>2</sup>	I <sub>S,pulse</sub>	T <sub>C</sub> =25°C			208	А
Forward voltage	V <sub>FSD</sub>	V <sub>GS</sub> =0V, I <sub>F</sub> =20A, T <sub>J</sub> =25°C		1.14	1.46	V
Torward voltage	▼ FSD	V <sub>GS</sub> =0V, I <sub>F</sub> =20A, T <sub>J</sub> =175°C		1.35		·
Reverse recovery charge	Q <sub>rr</sub>	V <sub>DS</sub> =400V, I <sub>S</sub> =50A, V <sub>GS</sub> =-0V, R <sub>G_EXT</sub> =50Ω		125		nC
Reverse recovery time	t <sub>rr</sub>	di/dt=1400A/μs, Τ <sub>J</sub> =25°C		12.5		ns
Reverse recovery charge	Q <sub>rr</sub>	$V_{DS}$ =400V, I <sub>S</sub> =50A, V <sub>GS</sub> =-0V, R <sub>G_EXT</sub> =50 $\Omega$		128		nC
Reverse recovery time	t <sub>rr</sub>	di/dt=1400A/µs, T_J=150°C		14.4		ns





### Typical Performance - Dynamic

Parameter	Symbol Test Conditions		Value			Units
Parameter	Symbol	lest Conditions	Min	Тур	Max	Units
Input capacitance	C <sub>iss</sub>	- V <sub>DS</sub> =400V, V <sub>GS</sub> =0V -		1414		
Output capacitance	C <sub>oss</sub>	$v_{DS} = 400 \text{ v},  \text{v}_{GS} = 0 \text{ v}$ = f=100kHz		118		pF
Reverse transfer capacitance	C <sub>rss</sub>	1-100KHZ		2		
Effective output capacitance, energy related	C <sub>oss(er)</sub>	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		150		pF
Effective output capacitance, time related	C <sub>oss(tr)</sub>	$V_{DS}$ =0V to 400V, $V_{GS}$ =0V		280		pF
C <sub>OSS</sub> stored energy	E <sub>oss</sub>	V <sub>DS</sub> =400V, V <sub>GS</sub> =0V		12		μJ
Total gate charge	$Q_{G}$	– V <sub>DS</sub> =400V, I <sub>D</sub> =50A, –		37.8		
Gate-drain charge	$Q_{GD}$	$V_{DS} = 400 \text{ V}, \text{ I}_{D} = 50 \text{ A},$ - $V_{GS} = 0 \text{ V} \text{ to } 15 \text{ V}$		8		nC
Gate-source charge	$Q_{GS}$	VGS 0V 1015V		11.8		
Turn-on delay time	t <sub>d(on)</sub>	Note 4,		13		
Rise time	t <sub>r</sub>	V <sub>DS</sub> =400V, I <sub>D</sub> =50A,		23		
Turn-off delay time	$t_{d(off)}$	Gate Driver = 0V to +15V, Turn-on $R_{G,EXT}=1\Omega$ ,		136		ns
Fall time	t <sub>f</sub>	Turn-off $R_{G,EXT}$ =50 $\Omega$		17.6		
Turn-on energy	E <sub>ON</sub>	Inductive Load, FWD: same device with		209		
Turn-off energy	E <sub>OFF</sub>	$V_{GS} = 0V, R_G = 50\Omega,$		212		μJ
Total switching energy	E <sub>TOTAL</sub>	TJ=25°C		421		
Turn-on delay time	t <sub>d(on)</sub>	Note 4,		10.5		
Rise time	t <sub>r</sub>	V <sub>DS</sub> =400V, I <sub>D</sub> =50A,		26		
Turn-off delay time	$t_{d(off)}$	Gate Driver = 0V to +15V, Turn-on $R_{G,EXT}$ =1 $\Omega$ ,		146		ns
Fall time	t <sub>f</sub>	Turn-off $R_{G,EXT}$ =50 $\Omega$		20		
Turn-on energy	E <sub>ON</sub>	Inductive Load, FWD: same device with		245		
Turn-off energy	E <sub>OFF</sub>	$V_{GS} = 0V, R_G = 50\Omega,$		248		μJ
Total switching energy	E <sub>TOTAL</sub>	Т <sub>Ј</sub> =150°С		493		

4. Measured with the half-bridge mode switching test circuit in Figure 23.





### Typical Performance - Dynamic (continued)

Parameter	Symphol	Symbol Test Conditions		Value			
Parameter	Symbol	Test Conditions	Min	Тур	Max	- Units	
Turn-on delay time	t <sub>d(on)</sub>			19			
Rise time	t <sub>r</sub>	Note 5 and 6,		27			
Turn-off delay time	t <sub>d(off)</sub>	V <sub>DS</sub> =400V, I <sub>D</sub> =50A, Gate Driver =0V to +15V,		41.6		ns	
Fall time	t <sub>f</sub>	$R_{G,EXT} = 1\Omega$ , inductive Load,		10.4			
Turn-on energy including R <sub>s</sub> energy	E <sub>ON</sub>	FWD: same device with $V_{GS}$		169			
Turn-off energy including R <sub>s</sub> energy	E <sub>OFF</sub>	= 0V and $R_G = 1\Omega$ , RC snubber: $R_S = 10\Omega$ and		149		1	
Total switching energy	E <sub>TOTAL</sub>	$ \begin{array}{c} \text{Snubber: } \kappa_{s} = 1002 \text{ and } \\ C_{s} = 300 \text{pF}, \\ T_{j} = 25^{\circ}\text{C} \\ \end{array} $		318		μJ	
Snubber R <sub>s</sub> energy during turn-on	E <sub>RS_ON</sub>			5			
Snubber R <sub>s</sub> energy during turn-off	E <sub>RS_OFF</sub>			8.5			
Turn-on delay time	t <sub>d(on)</sub>			17			
Rise time	t <sub>r</sub>	Note 5 and 6,		29			
Turn-off delay time	t <sub>d(off)</sub>	$V_{DS}$ =400V, $I_D$ =50A, Gate Driver =0V to +15V.		41		ns	
Fall time	t <sub>f</sub>	$R_{G,EXT} = 1\Omega$ , inductive Load,		9			
Turn-on energy including R <sub>s</sub> energy	E <sub>ON</sub>	FWD: same device with $V_{GS}$		198			
Turn-off energy including R <sub>S</sub> energy	E <sub>OFF</sub>	= 0V and $R_G = 1\Omega$ , RC snubber: $R_S=10\Omega$ and $C_S=300pF$ , $T_J=150^{\circ}C$		153		1	
Total switching energy	E <sub>TOTAL</sub>			351		μJ	
Snubber R <sub>s</sub> energy during turn-on	E <sub>RS_ON</sub>			5		1	
Snubber R <sub>s</sub> energy during turn-off	E <sub>RS_OFF</sub>			7		1	

5. Measured with the switching test circuit in Figure 24.

6. In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.



### Typical Performance Diagrams

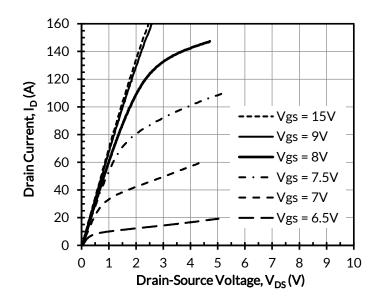


Figure 1. Typical output characteristics at T<sub>J</sub> = - 55°C, tp < 250 $\mu$ s

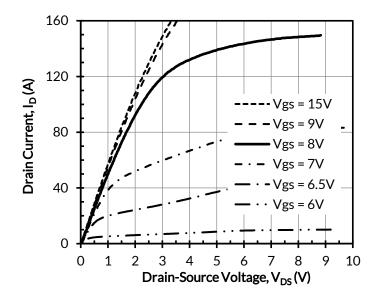


Figure 2. Typical output characteristics at  $T_J = 25^{\circ}C$ , tp < 250 $\mu$ s

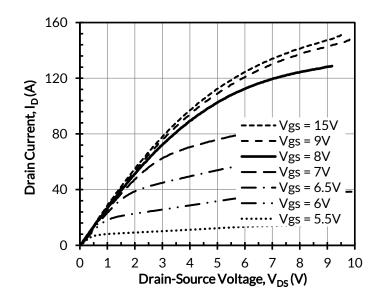


Figure 3. Typical output characteristics at  $T_J = 175^{\circ}C$ , tp < 250 $\mu$ s

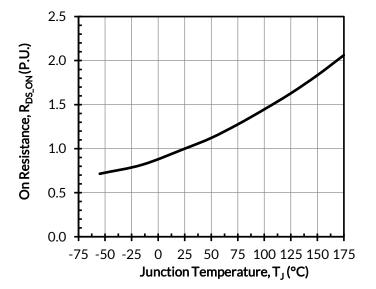


Figure 4. Normalized on-resistance vs. temperature at  $V_{GS}$  = 12V and  $I_D$  = 50A

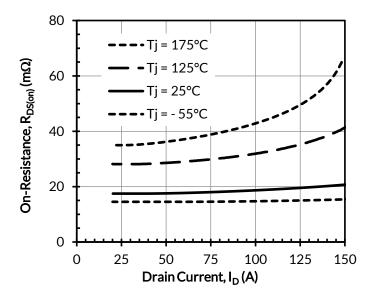
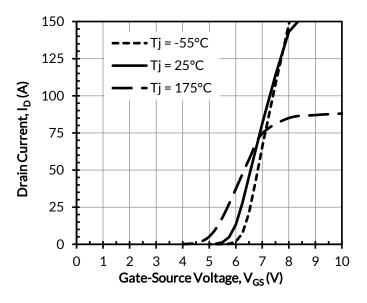


Figure 5. Typical drain-source on-resistances at  $V_{GS}$  = 12V



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Figure 6. Typical transfer characteristics at  $V_{DS}$  = 5V

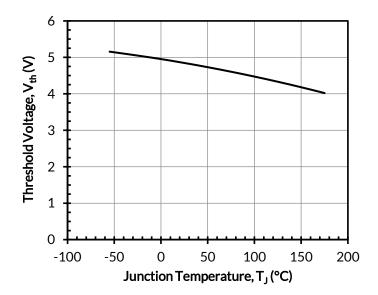


Figure 7. Threshold voltage vs. junction temperature at  $V_{\text{DS}}$  = 5V and  $I_{\text{D}}$  = 10mA

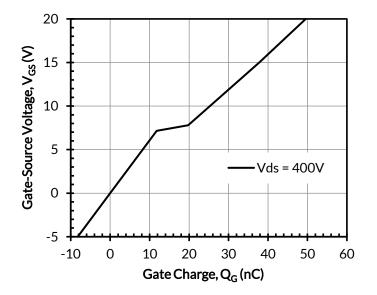


Figure 8. Typical gate charge at  $I_D$  = 50A

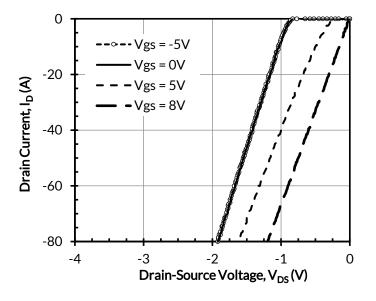


Figure 9. 3rd quadrant characteristics at  $T_J = -55^{\circ}C$ 

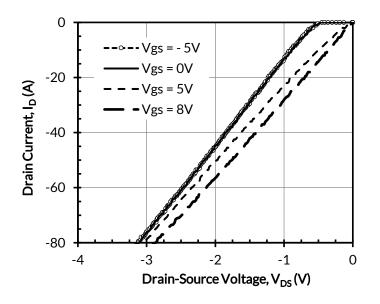
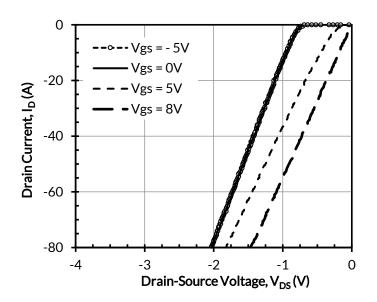


Figure 11. 3rd quadrant characteristics at T<sub>J</sub> = 175°C



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Figure 10. 3rd quadrant characteristics at  $T_J = 25^{\circ}C$ 

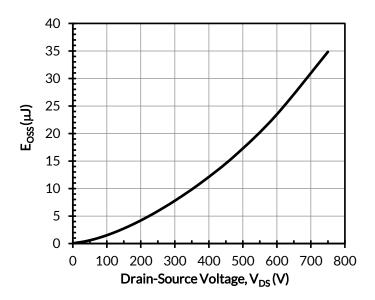


Figure 12. Typical stored energy in  $C_{OSS}$  at  $V_{GS}$  = 0V

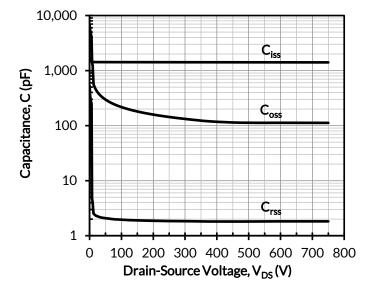
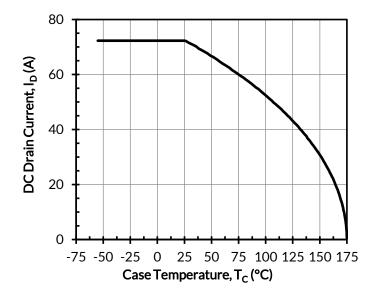


Figure 13. Typical capacitances at f = 100kHz and  $V_{GS}$  = 0V



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Figure 14. DC drain current derating

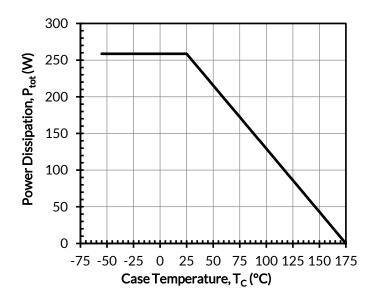


Figure 15. Total power dissipation

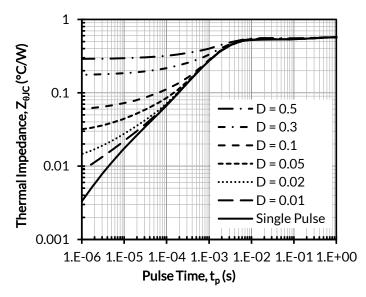


Figure 16. Maximum transient thermal impedance

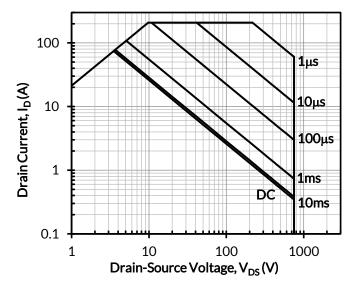
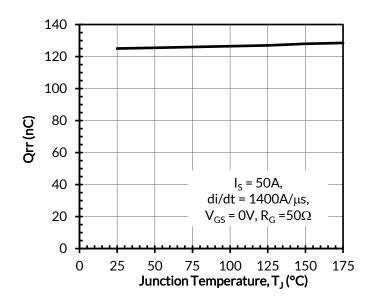


Figure 17. Safe operation area at  $T_{\rm C}$  = 25°C, D = 0, Parameter  $t_{\rm p}$ 



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Figure 18. Reverse recovery charge Qrr vs. junction temperature at Vds = 400V

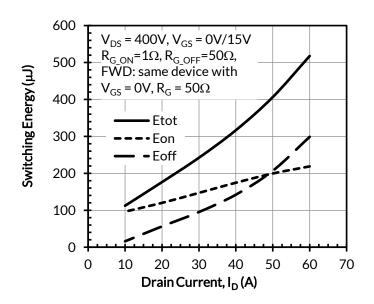


Figure 19. Clamped inductive switching energy vs. drain current at  $V_{DS}$  = 400V and  $T_J$  = 25°C

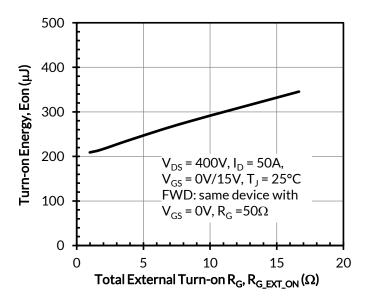


Figure 20. Clamped inductive switching turn-on energy vs.  $R_{G,EXT_ON}$ 



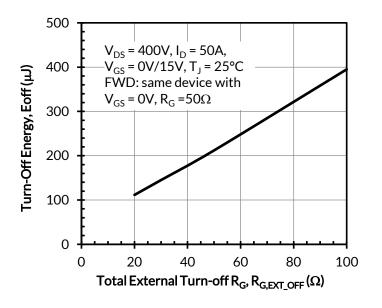
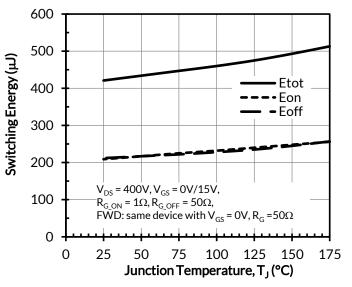


Figure 21. Clamped inductive switching turn-off energy vs.  $R_{G,\text{EXT\_OFF}}$ 



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Figure 22. Clamped inductive switching energy vs. junction temperature at  $V_{DS}$  =400V and  $I_D$  = 50A

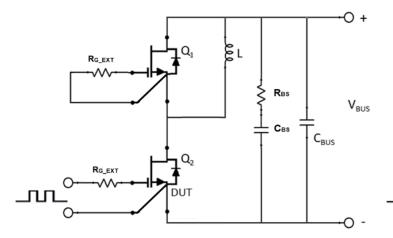


Figure 23. Schematic of the half-bridge mode switching test circuit. Note, a bus RC snubber ( $R_{BS} = 2.5\Omega$ ,  $C_{BS}=100$ nF) is used to reduce the power loop high frequency oscillations.

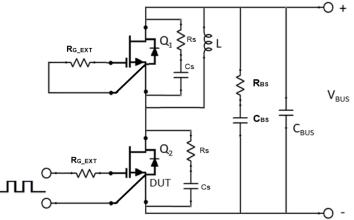


Figure 24.Schematic of the half-bridge mode switching test circuit with device RC snubbers (Rs =10 $\Omega$ , Cs = 300pF) and a bus RC snubber (R<sub>BS</sub> = 2.5 $\Omega$ , C<sub>BS</sub>=100nF).





### **Applications Information**

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ( $R_{DS(on)}$ ), output capacitance ( $C_{oss}$ ), gate charge ( $Q_G$ ), and reverse recovery charge ( $Q_{rr}$ ) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

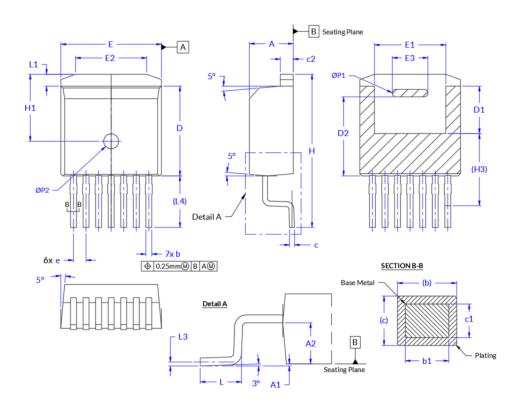
A snubber circuit with a small  $R_{(G)}$ , or gate resistor, provides better EMI suppression with higher efficiency compared to using a high  $R_{(G)}$  value. There is no extra gate delay time when using the snubber circuitry, and a small  $R_{(G)}$  will better control both the turn-off  $V_{(DS)}$  peak spike and ringing duration, while a high  $R_{(G)}$  will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high  $R_{(G)}$ , while greatly reducing  $E_{(OFF)}$  from mid-to-full load range with only a small increase in  $E_{(ON)}$ . Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com

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#### PACKAGE OUTLINE



	7L-D2PAK								
SYM	M	М	IN	СН					
SIM	Min	Max	Min	Max					
A	4.30	4.56	.169	.180					
A1	0.00	0.25	.000	.010					
A2	2.45	2.75	.096	.108					
b	0.50	0.70	.020	.028					
b1	0.50		.020	-					
с	0.40	0.60	.016	.024					
c1	0.40		.016						
c2	1.20	1.40	.047	.055					
D	8.93	9.23	.352	.363					
D1	4.65	4.95	.183	.195					
D2	7.90	8.10	.311	.319					
e	1.27	BSC	.050 BSC						
E	10.08	10.28	.397	.405					
E1	6.82	7.62	.269	.300					
E2	6.50	8.60	.256	.339					
E3	3.50	3.70	.138	.146					
н	15.00	16.00	.591	.630					
H1	6.68	6.88	.263	.271					
H3	7.3	REF.	.287	REF					
L	1.90	2.50	.075	.098					
L1	0.98	1.42	.039	.056					
L3	0.25	BSC	.0098	BSC					
L4	5.22	REF	.205	REF					
ØP1	0.65	0.85	.026	.033					
ØP2	1.40	1.60	.055	.063					

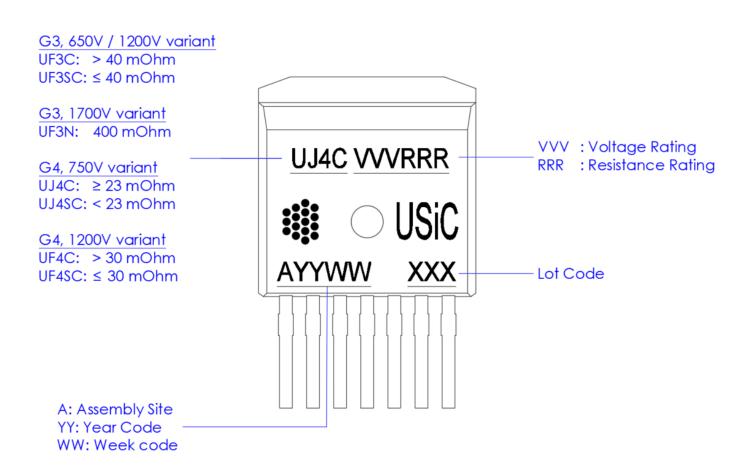
#### Notes:

- 1. GENERAL TOLERANCE: ±0.1 unless otherwise specified
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. PACKAGE BODY SIDES EXCLUDE MOLD FLASH AND GATE BURRS.
- 4. DIMENSION L IS MEASURED IN GAUGE LINE.
- 5. CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
- 6. DIMENSION c1 AND b1 APPLIES TO BASE METAL ONLY



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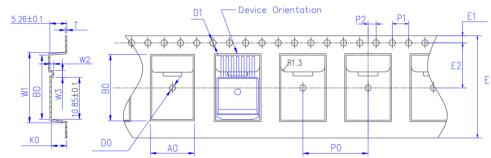
#### PART MARKING





#### PACKING TYPE

#### Carrier Tape

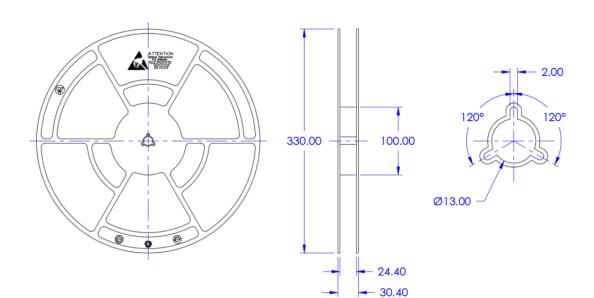


UNIT: MM

PACKAGE	AO	BO	KO	DO	D1	E	E1	E2	P0	P1	P2	Т
D2PAK (24 mm)	10.80 ±0.10	16.30 ±0.10	4.70 ±0.10	1.50 ±0.10	1.50 +0.1 -0	24.00 ±0.30	1.75 ±0.10	11.50 ±0.10	16.00 ±0.10	4.00 ±0.10	2.00 ±0.10	0.35 ±0.10

Exte	erior	size	
-	W1	16.9±0.1	
Spec 1	W2	1.3±0.1	
	W3	1.0±0.1	
-	W1	17.2±0.1	0
Spec 2	W2	1.8±0.1	Б
2	W3	0.85±0.1	$\bigcirc$

<u>Reel</u>



All dimensions in millimeters Anti-Static Tape and Rell (T&R) Quantity per Reel: 800 units



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#### **REVISION HISTORY**

Revision	Create Date (mm/dd/yyyy)	Description of Change	Initiator of Change
С	11/06/2023	Updated to Qorvo template Updated Package outline drawing based latest drawing revision	Glenn Galang
D	05/21/2024	Added illustration of device orientation on carrier tape (page 3)	Glenn Galang

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